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JPRS L/9806 24 June 1981

## **USSR** Report

ELECTRONICS AND ELECTRICAL ENGINEERING

(FOUO 7/81)



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JPRS L/9806

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# USSR REPORT ELECTRONICS AND ELECTRICAL ENGINEERING (FOUO 7/81)

#### CONTENTS

TELEMECHANICS, TELEMETERING, MACHINE DESIGNING AND PLANNING	
Applications of Computer Equipment and Automation in Electrical Power Systems	1
CERTAIN ASPECTS OF PHOTOGRAPHY, MOTION PICTURES AND TELEVISION	
Photoelectric Characteristics of the LI459 Vidicon	10
Color Television Vidicons With a Lead Oxide Target	16
COMMUNICATIONS, COMMUNICATION EQUIPMENT, RECEIVERS AND TRANSMITTERS, NETWORKS, RADIO PHYSICS, DATA TRANSMISSION AND PROCESSING, INFORMATION THEORY	
Evaluation of the Overall Loss of Trunking Circuits and Recording Circuits of Municipal Telephone Networks	26
Line Circuit Equipment of the K-1920P Transmission System	32
PUBLICATIONS, INCLUDING COLLECTIONS OF ABSTRACTS	
Abstracts From the Collection 'MODERN METHODS OF SIGNAL PROCESSING'	41
Electrical Engineering Handbook (in Three Volumes): 'VOLUME 1; GENERAL ASPECTS. ELECTRICAL ENGINEERING DATA'	48
Electrical Power-Supply Sources With Special Characteristics	62
Experimental Studies on Radio Wave Propagation	71
Fundamentals of the Electromagnetic Method of Stress Analysis in Anisotropic Media	74

- a - [III - USSR - 21E S&T FOUO]

Handbook of Shipboard Radio Communication and Radio Navigation Equipment. Vol. 2: Radio Navigation Equipment	76
Measurement of Electric Power in the Audio Frequency Range	80
Precision Design in Microelectronics	82
Reliability of Multichannel Communications Systems	84
Synthesis of Redundant Discrete Systems With Structural Reconfiguration	86
Theory and Calculation of Optoelectronic Instruments	89
Theory of Code Division of Signals	93
Transmission of Images in Digital Form	97

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CERTAIN ASPECTS OF COMPUTER HARD AND SOFT WARE: CONTROL, AUTOMATION, TELEMECHANICS, TELEMETERING, MACHINE DESIGNING AND PLANNING

UDC 621.311.621.316.621.314.629.396

APPLICATIONS OF COMPUTER EQUIPMENT AND AUTOMATION IN ELECTRICAL POWER SYSTEMS

Kiev PRIMENENIYE VYCHISLITEL'NOY TEKHNIKI I AVTOMATIZATSIYA V ELEKTROENERGETICHESKIKH SISTEMAKH in Russian 1980 (signed to press 14 Oct 80) pp 2, 215-224

[Annotation and abstracts of papers in the collection "The Application of Computer Equipment and Automation in Electrical Power Engineering Systems", editor-in-chief L.V. Tsukernik, Izdatel'stvo "Naukova dumka", Institute of Electrodynamics of the Ukrainian SSR Academy of Sciences, 800 copies, 224 pages]

[Text] Questions of the development of procedures, algorithms and programs (including comprehensive programs using a single data base) for the calculation of the normal and emergency modes of complex electrical power systems are treated in this collection. Urgent problems of relay protection as well as the design and study of automation and remote control devices are also analyzed.

The collection is intended for scientific, engineering, technical workers, electrical power engineers and specialists in the application of computers to electrical power engineering; it will also be useful to students of the higher educational institutes in electrical power engineering specialties.

#### Editorial Staff:

L.V. Tsukernik (editor-in-chief), V.N. Avramenko, K.V. Korobchuk, O.M. Kostyuk, V.A. Krylov, V.V. Rogoza, I.M. Sirota, B.S. Stogniy (deputy editor-in-chief), P.A. Chernenko (executive secretary) and V.N. Shestopalov.

UDC 621.311.011.1

A PROCEDURE FOR THE ANALYSIS OF LONG DURATION ELECTROMECHANICAL TRANSIENT PROCESSES IN A POWER SYSTEM IN THE CASE OF DISTURBANCES WHICH LEAD TO LARGE FREQUENCY CHANGES

[Abstract of paper by Avramenko, V.N. and Olyanishin, V.A.]

[Text] A procedure is proposed for the analysis of long duration electromechanical transient processes with significant frequency excursions; the possibilities for realizing it with the ChASTOTA-1 program are indicated and the results of calculations as well as comparisons with program analysis of the dynamic stability of the UDAR-2m are given. Figures 4; references 10.

1

UDC 621.311.1

AN ALGORITHM AND COMPUTATIONAL ROUTINE FOR THE STEADY-STATE MODES OF POWER SYSTEMS TAKING INTO ACCOUNT A SPECIFIED OR OPTIMALIZING FREQUENCY

[Abstract of paper by Sukhenko, V.I.]

[Text] A method is developed for the calculation of the steady-state modes of a power system, taking its frequency characteristics into account. Special attention is devoted to accelerating the rate of convergence of Newton's method in calculating the optimalizing frequency. The question of accounting for the available reactive power of synchronous generators is analyzed in the calculations of the steady-state modes of power systems. References 9.

UDC 621.316.925

THE CALCULATION ON THIRD GENERATION COMPUTERS OF THE OPERATING CONDITIONS INVOLVING INCOMPLETE PHASES IN COMPLEX ELECTRICAL NETWORKS

[Abstract of paper by Krylov, V.A. and German, A.I.]

[Text] An algorithm developed for the calculation of operating modes with incomplete phases in complex power grids is analyzed. The basis for the algorithm is the insertion of an additional voltage source in the switched branch in the original symmetrical network. The industrial program compiled for third generation computers in which the algorithm considered here is realized is described.

UDC 621.316.001.24

THE CALCULATION OF THE PRE-EMERGENCY CONDITIONS IN COMPLEX POWER SYSTEMS WITH A LARGE NUMBER OF NODES FOR THE SUBSEQUENT CALCULATION OF THE SHORT CIRCUIT CURRENTS CURRENTS

[Abstract of paper by Khrushchova, Ye.V. and Krylov, V.A.]

[Text] An algorithm for one of the variants of Newton's method is described as applied to calculations of the pre-emergency conditions in complex power systems. The computational results are intended for the subsequent calculation of the short circuit current. References 4.

UDC 621.311

THE DEVELOPMENT OF AN OPERATIONAL EXPERIENCE WITH COMPLEXES FOR DETERMINING THE EQUIVALENT CIRCUITS FOR THE CALCULATION OF STEADY-STATE OPERATING MODES

[Abstract of article by Kachanova, N.A., Kozlenko, A.N., Makarevich, R.A. and Perga, S.P.]

[Text] The following equivalent circuit design complexes are described: a) The "Ekvis-4", which is interfaced with the "Set'" program for steady-state operational mode calculations, developed by the TsDU [central dispatcher control] of the

2

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YeES [unified power system] of the USSR; b) The "Ekvis-5", which is interfaced with a variant of the "Set'" program for steady-state operational mode calculations, developed by the Institute of Electrodynamics of the Ukrainian SSR Academy of Sciences; c) The "Ekvis-6", which is interfaced with the "RUER" program for steady-state operating mode calculations, developed by the ODU Yug [southern integrated dispatcher control]. The volume of problems which can be solved: the original circuit - 1,000 nodes; the equivalent circuit region - 500 nodes. The complexes are written in FORTRAN-4 for unified series computers and the M-4030. Figures 4; tables 2; references 9.

UDC 621.311

A PROGRAM FOR SELECTIVELY DETERMINING EQUIVALENT CIRCUITS OF A NETWORK TO CALCULATE STEADY-STATE MODES

[Abstract of paper by Kachanova N.A. and Kozlenko A.N.]

[Text] A program for selectively determining an equivalent circuit is described, which includes the following operations: the transformation of parallel lines, the exclusion of radial lines and the exclusion of nodes having two ties to adjacent nodes. An example of a calculation using the program is given. The program is written in FORTRAN-4 for the unified series of computers and the M-4030. Figures 2; tables 3.

UDC 621.311:001.33:001.57

THE GENERATION OF A LIBRARY OF CALCULATED STEADY-STATE OPERATIONAL MODES FOR ELECTRICAL POWER SYSTEMS

[Abstract of paper by Chernenko, P.A. and Zorin, V.V.]

[Text] The construction of a library of operational modes which have been calculated beforehand (BR) is described. Methods of classifying the operational modes are treated, as well as the ascertaining of the significant criteria. The operating mode library is intended for the storage and output on displays of information on the steady-state mode based on a specified complete or limited set of significant criteria and can also be used in the operation and plannning of electrical power systems. References 4.

UDC 621.391.2:621.311.1

THE SPECIALIZED IVS-IED INFORMATION COMPUTER SYSTEM

[Abstract of article by Zharkin, V.F., Kuz'min, I.A. and Rudenko, P.N.]

[Text] The general concepts of the development of the specialized IVS-IED information computer system are proposed. The requirements placed on specialized information computer systems are formulated. The principles of program dispatcher control of the information computer process are worked out based on the input language for the control of the information computer system operation. Figures 1; references 3.

3

UDC 621.311.001

THE STRUCTURE OF A DATA BASE FOR THE SOLUTION OF ELECTRICAL ENGINEERING PROBLEMS

[Abstract of article by Kosterev, N.V., Tsyganov I.V. and Tseplyayev, V.G.]

[Text] A formal description is given for the information file of a data base. Its logic structure is presented. The universal nature of its applications is demonstrated. Figures 2; references 5.

UDC 621.311.001.57

QUESTIONS IN THE DEVELOPMENT AND DEBUGGING OF MULTIPHASE PROGRAMS (USING THE EXAMPLE OF THE DEVELOPMENT OF A COMPREHENSIVE PROGRAM FOR THE ANALYSIS OF THE STATIC STABILITY OF COMPLEX POWER SYSTEMS)

[Abstract of article by Korobchuk, K.V., Sambur, S.B. and Khimyuk, I.V.]

[Text] Experience with the organization of the programming, interfacing and comprehensive debugging of large multiphase programs is brought to light. The methods employed by the authors in the development and debugging of a comprehensive program using a third generation computer to analyze the statistical stability of complex power systems, the STATUS program, is described; this program incorporates a number of algorithmically complex and voluminous routines (in terms of the program realization and the large amount of input and intermediate data which is used), which are oriented towards the practically complete utilization of the immediate-access memory of the computer in the case of autonomous use. The solutions employed by the authors in interfacing the programs written in different programming languages are set forth (FORTRAN-4 and the ASSEMBLER language). Figures 2; references 5.

UDC 621.391.2:621.311.1

AN ADAPTATION TECHNIQUE FOR THE INFORMATION MODELS OF POWER SYSTEMS

[Abstract of paper by Zharkin, V.F. and Rudenko, P.N.]

[Text] The operational mode for making corrections and changes in the parameters of power system models is described. A definition is given for the adaptation for information models of power systems. A generalized adaptation algorithm is proposed based on an information retrieval system using standardized data structures in the information models. Figures 1; references 2.

UDC 621.391.2:621.311.1

THE DESIGN OF A UNIT TO INTERFACE THE "STATUS" COMPREHENSIVE PROGRAM FOR THE ANALYSIS OF STATIC STABILITY TO THE SPECIALIZED IVS-IED INFORMATION COMPUTER SYSTEM

[Abstract of article by Trokhimenko, A.I.]

[Text] One of the variants of a unit for interfacing the program for static stability analysis to an information computer system (the IVS-IED) is proposed. Its algorithm is described. The advantages of running the "STATUS"O" program in conjunction with the IVS-IED are demonstrated. Figures 1; references 3.

UDC 621.311.016.35

ON THE DEGENERATION OF THE CHARACTERISTIC EQUATION OF MATHEMATICAL MODELS AND THE PROBLEM OF POWER SYSTEM STABILITY

[Abstract of article by Kostyuk, O.M.]

[Text] The question of the degeneration of the characteristic equation of a power system is discussed where this dengeneration occurs as a result of neglecting small parameters. It is emphasized that the degenerate equation solves the problem of original system stability and is therefore of practical importance, if the tending of the small parameters to zero carries away part of the roots of the total (nondegenerate) equation to infinity to the left of the imagined axis. However, if a portion of the roots is shifted to the imaginary axis or to a small neighborhood of it and there are no right roots, which are sufficiently remote from the imaginary axis, then the degenerate equations of the stability problem is not solved (a special case similar to Lyapunov's special case). It is shown that such is the characteristic equation of the positional model of a power system. References 12.

UDC 621.313.322

A STUDY OF A METHOD OF MATHEMATICALLY PROCESSING THE RESULTS OF AN EXPERIMENT TO DETERMINE THE TRANSIENT PARAMETERS OF SYNCHRONOUS GENERATORS

[Abstract of article by Tverdyakov, V.V.]

[Text] Likelihood criteria are analyzed, which limit the order of the mathematical model of a synchronous generator derived on the basis of the results of specially organized experiments. Figures 1; tables 1; references 9.

UDC 621.317

THE APPLICATION OF GENERAL PURPOSE COMPUTERS TO THE MEASUREMENTS OF ELECTRICAL POWER LOSSES IN 750 KV NETWORKS

[Abstract of article by Neyman, V.A. and Kuznetsov, G.G.]

[Text] A method and an algorithm for the estimation of electrical power losses in 750 KV networks using general-purpose computers are presented, which make it possible to reduce the estimate error by an order of magnitude as compared to the error of the existing estimation technique. The economic impact on the international 750 KV electrical power transmission line from the introduction of the proposed algorithm is discussed. Figures 2.

UDC 621.311:621.3.016.312:621.372

POWERING THE RAPIDLY CHANGING LOADS OF AN AUTONOMOUS POWER SYSTEM THROUGH SYMMETRICAL COMPONENT FILTERS

[Abstract of article by Vasetskiy, V.M. and Tsukernik, L.V.]

[Text] A method of powering a rapidly changing load through symmetrical component filters is described, in which the impact of the symmetrical rapidly changing load on the power supply operating conditions is practically eliminated, something which boosts the operational reliability as well as the electrical quality in independent power systems. Figures 5; references 4.

UDC 621.314.224

ON THE CHOICE OF THE OPERATING MODE OF AN ELECTROMAGNETIC CURRENT CONVERTER IN RELAY PROTECTION AND AUTOMATION DEVICES

[Abstract of article by Stogniy, B.S. and Slyn'ko, V.M.]

[Text] The amplitude-frequency and phase-frequency characteristics of an electromagnetic measurement current transducer (IPT) are derived, where the transducer operates as a current transformer and a transreactor. The change in the current and angular errors of the IPT is analyzed for the case of a change in the temperature conditions, and the promise of the application of a transreactor to power relay protection and automation devices is demonstrated. Figures 6; references 7.

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UDC 621.314.224

ON THE QUESTION OF THE CRITERIA FOR THE EVALUATION OF THE OPERATION OF CURRENT MEASUREMENT TRANSDUCERS IN RECTIFIER CONVERTER CIRCUITS

[Abstract of article by Chernenko, V.A. and Vaynshteyn, Yu.A.]

[Text] Expressions are derived which establish the relationship between the quantities characterizing the operation of a measurement current transducer in the steady-state mode. A conclusion is drawn concerning the possibility of evaluating the operation of a measurement current transducer based on only one criterion. Figures 3; tables 1; references 2.

UDC 621.314.224

AN ALGORITHM TO GENERATE EQUATIONS FOR THE CIRCUIT MODEL OF CURRENT MEASUREMENT TRANSDUCERS

[Abstract of article by Stogniy, B.S., Kifilenko, A.V. and Levitskiy, V.G.]

[Text] The major standard components are determined which are the basis for the design of electromagnetic current measurement transducers (IPT). A technique is proposed and an algorithm is described for the generation of the circuit model of a measurement current transducer, which utilizes information on its standard components and makes it possible to study devices of any configuration which are treated here. Figures 1; tables 1; references 8.

UDC 621.314.224

THE DETERMINATION OF TOLERANCES IN THE PRODUCTION OF LINE CURRENT TRANSFORMERS

[Abstract of article by Stogniy, B.S. and Orobets, Yu.N.]

[Text] Design expressions are derived for the determination of the ultimately permissible mean square deviations and tolerances of the major production parameters of line current measurement transducers. An example of the calculation of the tolerances of a single stage line current transformer is given. Figures 4; tables 1; references 6.

UDC 631,316,925

THE PULSE-WIDTH MODULATOR IN THE INFORMATION TRANSDUCERS OF DIGITAL RELAY PROTECTION DEVICES

[Abstract of article by Kholodenko, Yu.N.]

[Text] The requirements placed on a pulse-width modulator as an element in an information transducer for digital relay protection devices are analyzed. A modulator circuit is proposed. A number of the parameters and properties of the modulator are also analyzed, in particular, the conversion error. Figures 5; references 8.

7

UDC 621.398:621.391.3

AN ANALYSIS OF THE OPERATING MODES OF A TRANSMITTER OF A REMOTE CONTROL SYSTEM TRANSMITTING VIA DISTRIBUTION NETWORKS

[Abstract of article by Koshman, V.I.]

[Text] The operation of a transmitting device, designed in the form of a flipflop with an active inductive load in the legs of the circuit is analyzed and the conditions for its optimal operation are ascertained. Recommendations are made concerning the selection of the basic parameters of the flip-flop. The efficiency of the transmitter is determined. Figures 4; references 2.

UDC 621.398:621.391.3

A STUDY OF VOICE-FREQUENCY INTERFERENCE IN THE DISTRIBUTION NETWORKS OF AN IRRIGATION COMPLEX

[Abstract of article by Koshman, V.I. and Trach, I.V.]

[Text] The results are given for an experimental study of the harmonics and pulse noise occuring in the distribution network of an irrigation complex for various operational conditions of the complex. Figures 4; tables 2; references 6.

UDC 621.317.08

A METER FOR THE TRIGGERING ANGLES OF THE RECTIFIERS OF THE TRANSMITTER OF AN EMERGENCY CIRCULAR LOAD RELIEF SYSTEM

[Abstract of article by Buslov, L.I., Osnach, A.M. and Shestopalov, V.N.]

[Text] A meter for the triggering angles of the rectifiers of the transmitter of an emergency circular load relief system is described, in which standard devices are used to measure the time intervals. Figures 3.

UDC 621.316.925

A STATISTICAL METHOD FOR CALCULATING THE DAMAGE TO AN ELECTRICAL SYSTEM BY FALSE RESPONSES OF REALY PROTECTION FOR POWER TRANSMISSION LINES

[Abstract of article by Karinskiy, Yu.I.]

[Text] A method is described for calculating the damage to an electrical system due to excessive actuation and failures in the actuation of the relay protection for the electrical power transmission lines as applied to the problem of optimizing the settings of the measurement devices for remote and current protection circuitry. This damage depends on the increase in the probabilities of a disruption of the power supply to system consumers because of the false reactions of relay protection, taking into account preventive maintenance and post-breakdown repairs to the

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lines and equipment of substations as well as the cost of a disruption in the electrical power supply for each consumer, taking into account violations in the production process, rejects and insufficient product output. Figures 3; tables 3; references 7.

UDC 621.316.925

ON THE QUESTION OF THE SENSITIVITY OF PROTECTION AGAINST SHORTS TO GROUNDS OF POWER GENERATION UNITS

[Abstract of article by Bogachenkov, A.Ye.]

[Text] The specific features of the design of various kinds of protection against short circuits to ground in the generators of power units are treated, as well as their sensitivity to various values of the resistance of the neutral circuit, and transient resistances at the point of the short to ground. Figures 3; tables 1; references 16.

UDC 621.316.92.00124

THE CALCULATION OF THE IMPACT OF OVERVOLTAGES ON LIMITERS IN NETWORKS WITH AN INSULATED NEUTRAL

[Abstract of article by Samoylovich, I.S.]

[Text] A technique is presented for the statistical calculation of the effect of overvoltages on limiters in networks with an isolated neutral. The functions and parameters of the distribution of the major characteristics of the discharge current are determined, where there are data on the distribution of the overvoltage levels and the network parameters. Figures 3; tables 2; references 2.

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### CERTAIN ASPECTS OF PHOTOGRAPHY, MOTION PICTURES AND TELEVISION

UDC 621.385.832.564.4

PHOTOELECTRIC CHARACTERISTICS OF THE L1459 VIDICON

Moscow TEKHNIKA KINO I TELEVIDENIYA in Russian No 1, Jan 81 pp 43-46

[Article by A.Ye. Gershberg, N.N. Mikhaylov-Teplov, S.P. Nizhegorodov and Yu.V. Savel'yev]

[Text] The development of vidicons with targets based on cadmium selenide (cadmicons or halnicons) has substantially improved vidicon parameters. The sensitivity and signal uniformity over the image field have been increased, while the response lag and dark currents have been reduced. Along with this, the linearity of the rise in the photoelectric current, which is inherent in vidicons with such targets, has brought about a rapid rise in the charge accumulated on the target with the presence of illumination. Since the current of the commutated beam must be limited to assure high resolution of the device, the light range of the vidicon proves to be narrow. This circumstance forces increased complexity on the transmitting camera design through the introduction of light controlling devices. For this reason, it was desirable to design a vidicon which has the advantages of vidicons with cadmium selenide targets, the light characteristic of which would be nonlinear in a definite range ( $\gamma < 1$ ). The latter would provide for a slower rise in the charge accumulated on the target as a function of the light and a correspondingly wider light characteristic.

The LI459 vidicon is the first vidicon with a cadmium selenide target having an expanded light range. This is a half-inch vidicon with mixed beam control (electrostatic focusing and magnetic deflection), which is similar in terms of its structural design to the LI437 and LI448 vidicons, differing from the LI430 and LI453 vidicons only in the structural design of the grid positioned at the target. The parameters of the device are checked for a working target surface of 4.9 x 6.5 mm<sup>2</sup>, and a target illumination of 1 lux (a type "A" source) with standard TV resolution. Under such conditions, the vidicon provides for a signal current of no less than 0.1 µA, a dark current of no more than 5 nA, a resolving power at the center of no less than 500 lines, and no less than 400 lines at the corners, a residual signal 40 msec after the light is cut off of no more than 20% and a signal nonuniformity over the raster of no more than 15%. The set of parameters indicated here, as well as the spectral response, are extremely close to the parameters of cadmium selenide target vidicons (for example, the LI448). The difference consists in a slight increase in the sensitivity and the response lag. However, the light range of the latter is quite limited (for example, the minimum range of the LI448 is 3 lux). The LI459 vidicon is likewise tested at an illumination of 500 lux, at which it should provide a

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signal current of no less than 0.1 µA and have a resolution of no less than 500 lines.

Along with this, the parameters cited do not contain any information on the properties of the vidicons for illuminations in a range of 1 to 500 lux. Obtaining data on the properties of the devices throughout the entire range of illumination variation was the major task of this paper.

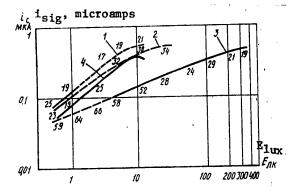


Figure 1.

The light characteristics of one sample of a LI459 vidicon. The numbers designate the residual signal after 40 msec. The regions of rise in the residual signal with illumination are shown by the dashed lines, while the regions of decay are shown by the solid lines.

U<sub>sp</sub>: 1. 30 volts;

2. 20 volts;

3. 10 volts;

4. The light characteristic of a cadmicon.

The light characteristics of one of the devices, which are shown in Figure 1, were measured using a slot in an opaque diaphragm. This provided for signal measurement in accordance with its definition as the difference in the currents at the black--white transition. Such a test pattern corresponds to natural subjects with a small illuminated area. The light characteristics were recorded for several values of the signal plate potential  $[{\tt U}_{\tt Sp}]$ . In this case, the greatest value of  ${\tt U}_{\tt Sp}$  corresponded to the optimal alignment for an illumination of 1 lux. At this light level and the maximum  $\textbf{U}_{\text{Sp}}, \ \gamma$  was close to unity. The light range for such values of  $\textbf{U}_{\text{Sp}}$  is 1 to 10 lux (depending on the specific value of  $\mathbf{U}_{\text{Sp}}$  and the properties of the sample). With an increase in the illumination and a reduction in  $\mathbf{U}_{\mathrm{Sp}}$ , the value of  $\gamma$  falls off. Thus, at the maximum  $U_{\mathrm{sp}}$ , the light characteristic is linear for low level illumination, and for high illumination, it is nonlinear. The light characteristic in cadmicons (see Figure 1) is linear under similar conditions practically to the point of image blooming, caused by insufficient beam current. At low values of  ${\tt U_{Sp}}$ and for signals which are not very small, the entire light characteristic is nonlinear. The appearance of nonlinearity provides for an expansion of the light range for a specified beam current and a specified value of the signal plate potential.

It must be added that the curve for the light characteristic will depend on the ratio of the transparent and opaque regions in the diapositive of the test pattern, since a consequence of the high target sensitivity is that a small fraction of the incident light which has passed through the target and is then reflected back to the target, substantially increases the dark current. For this reason, when a test pattern with considerably transparency is employed, the signal currents are lower and the light characteristic can change over from an increasing one to a falling characteristic, reducing the light range.

11

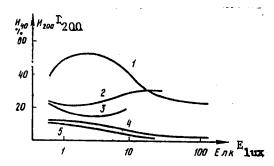


Figure 2. The change in the residual signal (I<sub>40</sub>, %; I<sub>200</sub>, %) in various regions of the light characteristic for one of the LI459 vidicons.

Key: I<sub>40</sub>: 1. 
$$U_{sp} = 10 \text{ v};$$
  
2.  $U_{sp} = 20 \text{ v};$   
3.  $U_{sp} = 30 \text{ v}, 40 \text{ v};$   
I<sub>200</sub>: 4.  $U_{sp} = 10 \text{ v};$   
5.  $U_{sp} = 20 \text{ v}, 30 \text{ v}, 40 \text{ v}.$ 

TABLE 1.

Pa	Napametp	S	omple Ospas			S	amp 1 e	2 4 13eu 4	
<sup>1</sup> 40	иА i <sub>c</sub> , мкА И <sub>40</sub> , % И <sub>к</sub> , %	0,1 30 6	0,2 40 10	0,3 47 8,5	0,1 30 3	0,2 33 3,5	0,3 36 8	0,4 38 7,5	0,5 50 6

TABLE 2.

.,	Isigna	al	ℓ <sub>C</sub> , MKĀ	μA		
bias	0.05	0,1	0,2	0,3	0,4	0,5
0 10 нА г	40 A 40	26 26	28 28	35 35	48 48	61 61
20 HA T		26 26	28	35 35	48	61

The lag (the residual signal 40 msec and 200 msec after the light is cut off) was measured at all points on the light characteristics of the vidicons. Three regions of residual signal levels are

observed on individual light characteristics: with an increase in the illumination from the minimal value, the response lag decreased, and thereafter increased, for the most part up to very high values (in particular, above those shown in Figure 2); it then dropped off, sometimes down to values substantially below the initial values. Some of the light characteristics contain two (rarely one) of the indicated regions. The first region (the reduction in the inertial lag) is characteristic of all vidicons with photodiode targets and is explained by the reduction in the switching lag with an increase in the target potential, which corresponds to an increase in the illumination and the signal.

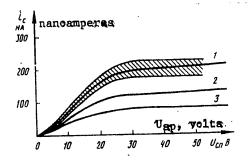


Figure 3. The volt-ampere characteristics of one of the LI459 samples.

Key: 1. E = 2 lux;

2. E = 1 lux;

3. E = 0.5 lux.

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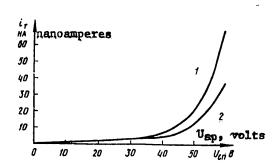


Figure 4. The dark current as a function of the voltage at the signal plate for two LI459 vidicon samples.

At reduced values of  $U_{\rm Sp}$ , the light characteristics now show  $\gamma < 1$  in the initial section; in this case, with an increase in the illumination, not a reduction, but rather an increase in the response lag is observed. The sharp increase in the response lag as a function of the illumination at values of the latter above the control point is a specific feature of LI459 vidicons.

Measurements of the commutation component of the response lag  $(I_k)$  were made with special equipment for a number of samples. The results of the measurements are shown in Table 1 for samples 3 and 4.

It follows from these data that for LI459 vidicons, the commutation lag amounts to an insignificant part of the overall response lag. It follows from this that the application of external bias lighting of the target to reduce the response lag is not expedient in the given case, something which is also illustrated by the values of the response lag (I40, %) which are shown in Table 2 (the bias lighting lag is characterized by the target current it produces:  $i_{bias}$ ).

The small value of the commutation response lag is explained in turn by the relatively small target capacitance (about 500 pF).

Typical volt-ampere characteristics and the dark current as a function of the signal plate voltage are shown in Figures 3 and 4 for two samples.

The spectral characteristics of LI459 vidicons, measured using the adopted procedure with standard test equipment, agree with the spectral characteristics measured under the same conditions for vidicons with a cadmium selenide target. In the case of measurements under typical conditions, the signal current is small, something which is due to the low energy sampled by the monochromator in a narrow spectral range.

To check how much the spectral characteristic determines the properties of a tube in all operating modes, the spectral characteristics were measured with changes in  $U_{\rm Sp}$  and the signal (energy) level (Figure 5). With a reduction in  $U_{\rm Sp}$  and the signal, the sensitivity in the shortwave portion of the spectrum fell off for all of the tested vidicons, all the way down to total insensivity to the blue and green portions of the spectrum. The maximum of the characteristic is shifted from approximately 700 to 720 nm. The change in the spectral characteristic with an increase in the illumination is greatest in a certain range of light levels, which is different for various samples.

To quantitatively estimate target burn, the central region of the target of an LI459 vidicon was illuminated for 15 minutes at  $i_s = 0.15~\mu\text{A}$  at an illumination of 1 lux, 10 lux and 100 lux (the signal plate voltage was varied). The illumination was accomplished through a hole in an opaque diaphragm, after which it was taken out and the burn signal (negative polarity) was immediately measured for the

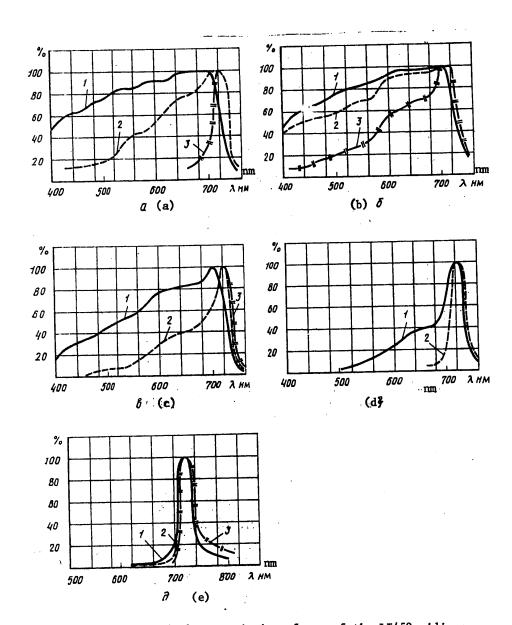


Figure 5. The spectral characteristics of one of the LI459 vidicon in various operating modes.

a.  $U_{\rm Sp} = 50$  volts;  $i_{\rm s}$ : 1. 30 nA; 2. 250 nA; 3. 500 nA; b.  $U_{\rm Sp} = 40$  volts;  $i_{\rm s}$ : 1. 30 nA; 2. 90 nA; 3. 150 nA; c.  $U_{\rm Sp} = 30$  volts;  $i_{\rm s}$ : 1. 30 nA; 2. 80 nA; 3. 200 nA; d.  $U_{\rm Sp} = 20$  volts;  $i_{\rm s}$ : 1. 30 nA; 2. 70 nA; e.  $U_{\rm Sp} = 10$  volts;  $i_{\rm s}$ : 1. 14 nA; 2. 17 nA; 3. 30 nA.

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#### TABLE 3

Target Burn Mode, lux		Burn Signal pamperes)
Mode, 1ux	Sample 1	Sample 2
1	0.002	0.016
10	0.011	0.017
100	0.002	0.002

uniformly illuminated field. The results of the measurements using the procedure described above are presented in Table 3.

It must be added that the target burn in the given target is of a reversible nature: it disappears after a few hours.

The light range in ordinary cadmicons can also be expanded in a manner similar to that presented above through a reduction in the signal plate potential. However, the target burn effect is manifest sharply in this case. Moreover, with a reduction in the signal plate potential, the signal at the corners drops out earlier for cadmicons (the corners "bloom").

#### Conclusions

- 1. The LI459 vidicon makes it possible to operate with control of the voltage at the signal plate in a wide light range while retaining satisfactory light parameters. In the maximum sensitivity mode (with a constant  $U_{\rm sp}$ ) and with adjustment of the signal plate potential, it has a greater light range than a cadmicon.
- 2. It is most expedient to use the LI459 vidicon in simple applied black and white TV cameras, operating in a broad range without special light control devices. Because of the considerable change in the spectral characteristic as a function of the operating conditions, it cannot be used in color television cameras.

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COLOR TELEVISION VIDICONS WITH A LEAD OXIDE TARGET

Moscow TEKHNIKA KINO I TELEVIDENIYA in Russian No 1, Jan 81 pp 38-42

[Article by G.S. Vil'dgrube, M.A. Kalantarov, V.A. Kozlov, A. G. Lapuk and O.A. Timofeyev]

[Text] Vidicons with a lead oxide target (OSM) are the main type of transmitting tubes in use throughout the world for color television broadcasting [1]\*. This is due to the following properties of vidicons with lead oxide targets:

--The relatively high sensitivity; the close match between the spectral sensitivity and the sight curve of the eye; the small scatter in the slope of the light characteristic from device to device; the low dark current level; the fast response; and the slight dependence of the parameters on temperature in a range of -3- to +50 °C.

The following type of lead oxide target vidicons are being series produced by domestic industry: LI432, LI442, LI460, LI462, LI457 and LI458. All of the devices have magnetic focusing and beam deflection. Some of the specific structural design features of the vidicons (Figure 1) are noted in Table 1.

The Lead Oxide Target

The good operational performance of vidicons is determined in many regards by the photoconductive target. It takes the form of a polycrystalline lead oxide film with a thickness of 12 to 20 micrometers, applied to a conducting

<sup>\*</sup> Vidicons with lead oxide targets have company names abroad: Plumbicon ("Philips") and Leadicon ("English Electric").

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transparent signal plate. The crystallites have a plate-light shape with dimensions of approximately 0.1 x 3 x 0.05 micrometers. The specific surface, determined by radioactive krypton adsorption, is 30 to 50 m $^2$ /g. The film porosity is 50 to 60 percent.

The target on the substrate side consists predominantly of the rhombic modification of PbO, where the crystallites in the [010] plane are oriented parallel to the substrate. The remainder of the film is formed by tetragonal crystallites with the [110] plane oriented parallel to the substrate. The ratio of the amount of the tetragonal to the rhombic forms is approximately 9:1.

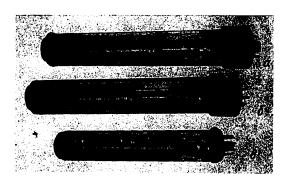


Figure 1.

An exterior view of the vidicons: top: LI432, LI442;

center: LI460, LI462; bottom: LI457, LI458. The energy structure of the target is specified by the fact that an intermediate layer with n-type conductivity is formed alongside the signal plate, and then enters a region with quasinatural conductivity (i), forming the bulk of the target thickness; then, a region with p-type conductivity is created by means at the target surface by means of special processing.

With the working voltage polarity (positive at the signal plate), the n and p-type intermediate layer play the part of the blocking contacts, and the entire system behaves in a manner similar to a pin-photodiode, connected in the cutoff direction. Because of this, as well as because of the considerable width

of the forbidden band of tetragonal lead oxide (1.9 eV), the dark current of the lead oxide target has an extremely small value. The dark current levels of vidicons, which fall in range of 0.5 to 3 nanoamperes, are due to scattered light and X-radiation, present inside the envelope of the device during the operation of its electron-optical system.

With a voltage of 45 V at the signal plate (the working voltage), a relatively high electrical field intensity is produced in the target, which provides for high efficiency in the separation of the electron-hole pairs generated by the light and the fast output of the charge barriers to the contact regions. At the working value of the voltage, the signal current is close to the saturated value, while the light-signal characteristic becomes close to a strictly linear response, and the slope of this characteristic amounts to  $0.95 \pm 0.05$  for various devices, something which makes it possible to easily accomplish the balancing of the

17

	Type of Electron Gun	Triode	Triode	Triode	Triode	Tetrode "anticomet"	The same	
rget Vidicons	Size of the Working Jarget Area, mm	17.1 x 12.8	$17.1 \times 12.8$	$17.1 \times 12.8$	$17.1 \times 12.8$	$12.7 \times 9.5$	$12.7 \times 9.5$	
f Lead Oxide Ta	Target Light Bias System	None	None	Yes	Yes	Yes	Yes	
Specific Design Features of Lead Oxide Target Vidicons	Spectral Sensitivity Range, nm	400 - 640	400 - 700	400 - 640	400 - 700	400 - 640	400 - 700	
Spec	Envelope Diameter (max), mm	30.6	30.6	30.6	30.6	26.2	26.2	
TABLE 1.	Type of Device	LI432	L1442	L1460	L1462		L1458	

18
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chrominance signals in the transmitting cameras without preliminary selection of the tubes according to the light characteristic.

The fast output of charge carriers to the contact regions leads to a small photoelectric inertia, while the porosity of the targets and their considerable thickness is responsible for relatively small capacitances of the working region (400 to 600 pF/cm<sup>2</sup>), and consequently, a reduction in the commutation component of the response inertia of the vidicons. Because of the combined effect of the factors indicated here, lead oxide target vidicons have the least response lag of the other types of vidicons.

The spectral characteristics of the vidicons are shown in Figure 2. The vidicons intended for operation in the blue channel have a characteristic which is shifted towards the shortwave end. All of the vidicons have a sensitivity maximum in a range of 450 - 525 nm. The quantum yield in this region figured on the basis of the absorbed light is close to unity. Beyond 550 nm, the sensitivity falls off rapidly; because of a reduction in the optical absorption in the lead oxide, the sensitivity practically disappears at 640 nm. This interferes with correct color transmission and in order to shift the red sensitivity boundary in the direction of longer wavelengths, sulfur is introduced into the target, which leads to the formation of oxysulfide compounds with lower photoelectric activation energies than for pure lead oxide. In this way, the red boundary is shifted and the sensitivity to red light is increased for the LI442, LI462 and LI458 vidicons.

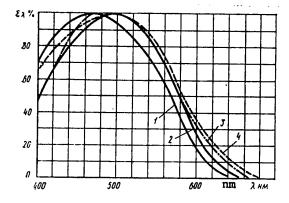


Figure 2. Vidicon spectral sensitivity characteristics.

Key: 1. For the blue channel, LI432 s, LI460 s, LI 457 s;

- For the brightness, green and red channels, LI432 ya, z, k; LI 460 ya, z, k; LI457 ya, z, k;
- LI458 ya, k [brightness, red channels];
- 4. LI442 ya, k; LI462 ya, k.

The Structural Design of Lead Oxide Target Vidicons

The specific features of the application of vidicons in multitube transmitting cameras place high requirements on the vidicons in terms of the value and identical nature of the geometric distortions. The scatter in the geometric distortions in a central region with a radius of 0.4h (where h is the image height) does not exceed 0.05%; in a region with a radius equal to half of the image width, it does not exceed 0.15% and in the remaining portion of the raster, does not exceed 0.3%. To meet these requirements, the envelopes of the vidicons are calibrated with respect to the inside diameters. The electron optics (EOS) are strictly centered relative to the envelope by means of two rigid clamps, made with small tolerances.

The position of the clamps can be seen in Figure 1. To reduce the scatter in the geometric distortions, the tolerances for the spacing between the grid and the target have been substantially reduced (as compared to other types of vidicons); using special assembly jigs, good coaxial alignment of the focusing electrode and the gun is achieved.

The rigid securing of the electron optics to the base seal and relative to the envelope, as well as the high resonant frequency for grid oscillations eliminate microphonics.

To reduce the response lag for LI460, LI462, LI457 and LI458 vidicons, internal bias lighting of the target is employed, which increases the signal readout efficiency for weakly illuminated portions of the image. In the first two vidicons, bias lighting is accomplished through a hole in the focusing electrode from an illuminator located outside the tube on its basing socket, and connected in parallel with the thermionic cathode heater. The bias lighting intensity is adjusted with a diaphragm placed in the illuminator.

In the LI457 and LI458 vidicons, the target is bias lighted by a lightguide with two outputs, arranged symmetrically inside the focusing electrode. The input end of the lightguide is sealed into the exhaust stem of the tube. The illuminating lamp is mounted in the vidicon power socket; the bias lighting intensity is adjusted by varying the voltage to the lamp.

TABLE 2 The Electrical Specifications for Lead Oxide Target Vidicons

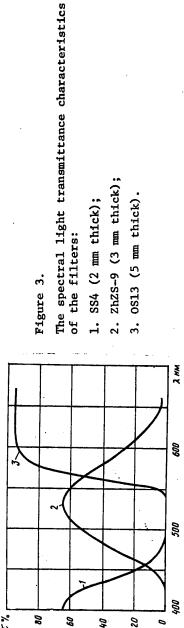
	L1432 L1442	LI460 LI462	L1457 L1458
Filament voltage, volts	6.3	6.3	6.3
Filament current without the illuminator, ma	300	300	. 90
Illuminator current, ma	, <del>-</del>	80	80
Voltage at the first anode, volts	300	300	300
Voltage at the supplemental clectrode, volts	_	_	300
Voltage at the second anode, volts	600	600	500
Grid voltage, volts	675	675	750

The internal surface of the focusing electrodes of all vidicons is given a matte finish to improve the target bias lighting uniformity. The nonuniformity in the target current, produced by the bias lighting, is symmetrical relative to the center of the target and does not exceed 15%.

Because of the increased diffuse scattering of light at lead oxide targets, all of the vidicons have anti-halo discs secured to the input window with an optical glue.

TABLE 3.	The	The Sensitivity and Modulation Level of Lead Oxide Target Vidicons	ity and	Modulati	on Level	of Lead	0x1d	Tar	get	Vid	cons	1
Parameter Designation	L1432ya L1460ya	LI432s LI460s		LI432z LI432k LI460z LI460k	LI442ya LI462ya	LI442k LI462k	LI457ya	ZLSTIT	s/5417	rids7k	LI458ya	rit28k
Sensitivity, µA/lm: minimal	360	40	140	100	450	160	360	130	04	75	130 40 75 400 115	115
typical	400	53	156	125	200	185	400 150 44	150	77	85	1.	130
Modulation level of the signal current in %:												
minimal	40	45	40	35	55	40	40	40	45	35	40 45 35 45 40	40
typical	44	26	20	42	09	20	74	44	20	ı	ı	45
[Suffixes break out as follows: "ya" = brightness; "s" = blue; "z" = green; "k" = red]	cout as	follows:	"ya" =	brightne	.82; "S"	= blue;	121	gre	en;	K	= rec	

Figure 3.



0S13 (5 mm thick).

21 FOR OFFICIAL USE ONLY

For vidicons with increased red sensitivity, an interference light filter with a primary passband in the 400 to 720 nm range is applied to the anti-halo disc to suppress the unnecessary sensitivity to light at a wavelength greater than 720 nm.

One of the special structural design features of the LI457 and LI458 vidicons is the "anticomet" (AK) tetrode type gun, which is distinguiahsed by an additional electrode in the cavity of the first anode.

During the beam retrace, when the horizontal line voltage pulses are being fed to the additional electrode, the modulator and the cathode accomplish the additional charging of the overilluminated (with an exposure of 200 - 300 lux) portions of the target with an increased beam current (100 - 150 microamperes), something which assures the transmission of the moving overilluminated portions of the image without characteristic defects in the form of trails resembling taffy pulls ("comet tails"). The action of the anticomet gun has been described in greater detail in [2].

A second special design feature of the LI457 and LI458 vidicons is the ceramic support ring with two local leads of the signal plate which are glued to the disc of the envelope. Such a structural design cuts the output capacitance of the vidicons by more than half (2.5 to 3 pF), which reduces the camera preamplifier noise.

The high precision in gluing the support ring with a coaxial misalignment of no more than 100 micrometers relative to the grid assembly provides for high precision in combining the vidicon rasters in multiple tube cameras.

Electrical Specifications, Major Light Performance Parameters and Characteristics

The overall size and connection dimensions of the LI432, LI442, LI460 and LI462 vidicons are the same. The LI457 and LI458 vidicons likewise have identical overall and connector dimensions. The electrical operating conditions of the devices are given in Table 2.

The sensitivity of lead oxide target vidicons (Table 3), because of the linearity of the light response, is constant in the working range of illuminations and is governed by the signal current level, referenced to the light flux falling on the target. The sensitivity is measured for the case of target illumination by a source with a color temperature of 2,850° K. When measuring the sensitivity of vidicons intended for chrominance channels, color filters are inserted in the path of the incident light: the SS-4 is for the blue channel, the ZhZS-9 is for the green and OS-13 is for the red channel. The transmittance characteristics of the light filters are shown in Figure 3.

The resolving power of lead oxide target vidicons exceeds 600 lines. The modulation level is determined at the 400 line mark (5 MHz) for the case of optimal focusing in the image field and a uniform frequency response of the measurement channel. The measurement of the modulation level and all of the other parameters, with the exception of the sensitivity is made for a modulator voltage which provides for the following signal readout:

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0.6 \mu\Lambda: for the LI432 ya, LI442 ya, LI460 ya, LI462 ya; 0.4 \mu\Lambda: for the LI457 ya, LI457 z, LI458 z; 0.3 \mu\Lambda: for the LI432 s, LI432 z, LI432 k, LI442 k, LI460 ya; 0.2 \mu\Lambda: for the LI457 s, LI457 k, LI458 k.
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The values obtained under the indicated conditions are given in Table 3.

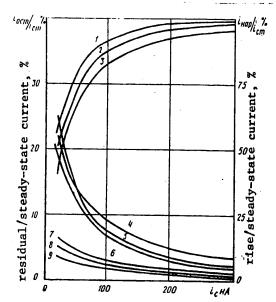


Figure 4. The signal rise and decay response lag.

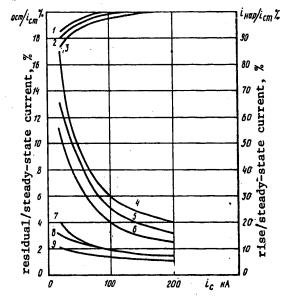


Figure 5. The signal rise and decay response lag

```
Key: 1. Rise (40 msec), LI432 s;
2. Rise (40 msec), LI442 ya, k;
3. Rise (40 msec), LI432 ya, z, k;
4. Decay (40 msec), LI442 ya, k;
5. Decay (40 msec), LI432 s;
6. Decay (40 msec), LI432 ya, z, k;
7. Decay (200 msec), LI432 s;
8. Decay (200 msec), LI442 ya, k;
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9. Decay (200 msec), LI432 ya, z, k.

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Key: 1. Rise (40 msec), LI457 ya, z;
2. Rise (40 msec), LI458 ya, z, k;
3. Rise (40 msec), LI457 s;
4. Decay (40 msec), LI458 ya, k;
5. Decay (40 msec), LI457 s;
6. Decay (40 msec), LI457 ya, z, k;
7. Decay (200 msec), LI458 ya, k;
8. Decay (200 msec), LI457 s;
9. Decay (200 msec), LI457 ya, z, k.
```

 $\Lambda11$  of the vidicons can operate at beam currents sufficient to read out signals down to 1.2  $\mu A,$  however, somewhat of a degradation in the modulation level and an increase in the geometric distortions as compared to the nominal beam current, are possible in this case.

As can be seen from Table 3, the modulation level is higher for vidicons for the blue channel and vidicons which have an increased sensitivity in the red portion of the spectrum. This is due to the fact that for lead oxide target vidicons, the factors which limit the resolving power are the light scattering in the target and

the spreading out of the potential relief on the commutated surface of the target. In the case of illumination with blue light, the scattering is small because of the strong absorption. For the LI442, LI462 and LI458 vidicons, the modulation level is greater because of the increase in the surface resistance of the target caused by the introduction of sulfur.

The most important parameter of vidicons is the response lag. For lead oxide target vidicons, the response lag for the signal rise and decay is governed, just as for other vidicons, by the ratio of the signal generated after a definite time has elapsed following the illumination being turned on or off to the steady-state value of the signal. This ratio is usually expressed in percent and is defined at 40 msec and 200 msec intervals, i.e., after the fourth or fifth frame following the change in illumination.

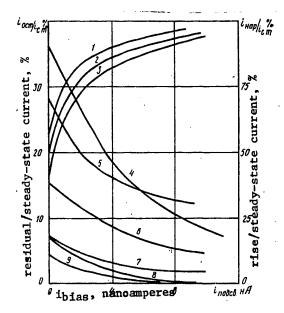


Figure 6. The response lag in the rise and decay as a function of the current produced by the bias lighting.

Key: 1. Rise (40 msec), LI460 ya, z, k;

- 2. Rise (40 msec), LI460 s;
- 3. Rise (40 msec), LI462 ya;
- 4. Decay (40 msec), LI462 ya, k;
- 5. Decay (40 msec), LI460 s;
- Decay (40 msec), LI460 ya, z, k;
- 7. Decay (200 msec), LI460 s;
- 8. Decay (200 msec), LI460 s;
- 9. Decay (200 msec), LI460 ya, z, k.

The curves for the signal response rise and decay lag as a function of the signal current level are shown in Figures 4 and 5. The lower values of the response lag for equal signal currents for the LI457 and LI458 vidicons, as compared to the LI432, LI460 and LI462 are due to the lesser capacitance of the working target surface.

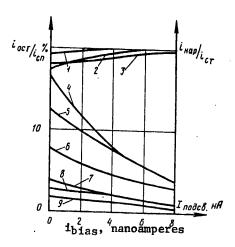


Figure 7. The response lag in the decay and rise as a function of the current produced by the bias lighting.

Key: 1. Rise (40 msec, 40 nA), LI457 ya, z;

- 2. Rise (40 msec, 20 nA), LI457 s;
- 3. Rise (40 msec, 20 nA), LI458 ya, k;
- 4. Decay (40 msec, 20 nA), LI458 ya, k;
- 5. Decay (40 msec, 20 nA), LI457 s;
- 6. Decay (40 msec, 40 nA), LI457 ya, z;
- 7. Decay (200 msec, 20 nA), LI458 ya, k;
- 8. Decay (200 msec, 20 nA), LI457 s;
- 9. Decay (200 msec, 40 nA), LI457 ya, z;

The curves for the signal current rise and decay lag are shown in Figures 5 and 6 as a function of the current produced by the bias lighting.

A substantial reduction in the lag occurs with an increase in the current produced by the bias lighting up to 5 nA (Figure 7). However, when cameras operate in reduced light, it is possible to increase the bias lighting intensity up to 8 - 10 nA in blue and red channel vidicons, where the lag is most pronounced. The bias lighting current nonuniformity does not exceed 15% in this case.

#### Conclusion

Two years of operational experience with the 30 mm diameter vidicons at 45 of the nation's television centers have demonstrated that the devices meet all of the modern requirements for color TV broadcasting during operation in three and four tube cameras.

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COMMUNICATIONS, COMMUNICATION EQUIPMENT, RECEIVERS AND TRANSMITTERS, NETWORKS, RADIO PHYSICS, DATA TRANSMISSION AND PROCESSING, INFORMATION THEORY

UDC 621.395.7

EVALUATION OF THE OVERALL LOSS OF TRUNKING CIRCUITS AND RECORDING CIRCUITS OF MUNICIPAL TELEPHONE NETWORKS

Moscow ELEKTROSVYAZ' in Russian No 11, Nov 80 pp 27-29

[Article by L. V. Andreyeva, M. Ya. Kaller, Yu. I. Kopacheva, Yu. A. Parfenov, and E. I. Sankovskiy, submitted 4 Jun 1979]

[Text] The automation of transmission processes of intercity telephone messages, introduction of crossbar ATS [automatic telephone exchanges] in local telephone networks, creation of multichannel transmission systems, cables of new types, and, at the same time, retention of the existing technical equipment, for example, coil-loaded circuits, made it necessary to evaluate telephone channels more precisely from the viewpoint of ensuring the quality of telephone communications and the possibility of transmitting service information by a multifrequency code via trunking circuits (SL) and recording circuits (ZSL) at the municipal telephone network.

This article gives the basic principles for evaluating overall attenuation of composite SL and ZSL which include low-frequency cables, KRR [cable radio relay] (KAMA) and IKM-30 transmission systems, bridge-type amplifiers, coil-loaded cables, and station quadripoles of various ATS with consideration for the mismatching of input resistances of the channel elements.

Methods of Constructing SL and ZSL and Their Elements. As is known, city telephone networks, with respect to their structure, are divided into single-office and multi-office networks, and the latter -- into networks with toll-switching planning and networks without toll-switching planning. The connection between rayon ATS (RATS), RATS and junction center ATS -- points of incoming messages (UVS) and points of outgoing messages (UIS), as well as between UVS and UIS, is accomplished through trunking circuits.

Automatic long-distance communication of subscribers of RATS with AMTS is accomplished through recording circuits, and in the opposite direction from AMTS to RATS -- through connecting lines of the long-distance trunk (SLM).

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The following two types of ATS are introduced widely in municipal telephone metworks of our country: ten-step-system ATS (ATS-DSh) and crossbar-type ATS (ATSK). In order to ensure automatic long-distance communication, domestic and foreign crossbar AMTS [automatic long-distance telephone exchanges] are introduced. The matching of the interaction of switching devices of various systems of ATS and AMTS is achieved by using sets of trunk line relays (RSL).

During the stage of establishing the connection of subscribers included in an ATS-DSh, the digital information of dialing and other interaction signals are transmitted through the SL by direct-current pulses, and in the case of the ATSK -- by a "2 out of 6" multifrequency code, i.e., pulses of two different frequencies out of six provided for in the code are delivered simultaneously into the line. Two-frequency signals arrive at the SL input from a multifrequency generator of the ATSK equipment. The output level of the signal is 7.4 dB. The working frequencies of the code are: 0.7, 0.9, 1.1, 1.3, 1.5 and 1.7 kHz. The same frequencies are used for transmitting signals of the equipment of automatic number identification (AON).

Digital information of the interaction signals between ATS and AMTS is transmitted through SLM and ZSL either by the multifrequency method of the above-mentioned frequencies, or with direct-current pulses by the decade method, depending on the type of the interacting ATS and AMTS [automatic long-distance telephone exchange].

In the speech channel, after establishing the connection via SL, ZSL and SLM, telephone information is transmitted in the analog form (frequency spectrum 0.3-3.4 kHz).

At an automatically switching city telephone network, composite sections of SL, ZSL, and SLM channels are formed by connecting several different elements: physical (non-multiplexed) cable lines; cable lines equipped with amplifiers; coil-loaded lines; channels of frequency and digital transmission systems; switching devices of junction ATS with interoffice connecting cables. In the designing of SL and ZSL (SLM), the choice of elements is determined by the necessity of ensuring the attenuation norms.

TG or TPP-type cables with cores 0.5 and 0.7 mm in diameter are used as physical cable lines for SL and ZSL (SLM). Their individual sections are equipped with UMT [bridge amplifiers] or are coil-loaded. In some instances, TZ-type cables with cores of 0.8 and 0.9 mm are introduced.

Trunk lines are organized through four-wire TCh [tone frequency] channels with the aid of KRR or KAMA transmission systems and matching equipment -- special multiplex trunk line relays (RSLU).

There are several types of RSLU used in SL and ZSL, depending on the type of the telephone exchange (ATS-DSh, ATSK, or AMTS), type of service (local or long distance), purposes of the sets (incoming or outgoing). Each set, regardless of its type, in addition to a relay, contains attenuators and differential systems included in the speech channel which determine the overall loss of the SL.

SL sections which have multichannel transmission systems can have a two or four-wire termination. For composite SL and ZSL (SLM) channels which include sections of coilloaded and noncoil-loaded physical cable circuits equipped with amplifiers (for uniform or mixed connection of the above-mentioned elements), the test jacks of the SL

27

distributing frames of RATS should be considered as an input or output. A composite SL, ZSL (SLM) formed by TCh channels of the KPP equipment has an input and output on two-wire differential systems (DS) located at terminal RATS.

Requirements for the Overall Loss of SL and ZSL Transmission Channels. These requirements are established on the basis of the condition of ensuring normal functioning of the speech channel in the spectrum of frequencies of 0.3-3.4 kHz. For the system of Statewide Automatically Switching Telephone Network (OAKTS), attenuation is distributed over the sections of the long-distance and local speech channel.

	Table 1		
(1) Назначение связи	(4) Участок тракта соединительных линий	Рабочее тухани кабель-	33.
ATC-ATC (2)	PATC-YBC (YMC) YBC-YHC (6) PATC-YBC-YHC- PATC (7)	13,0 9,0 18,0	7.0
ATC-\\\TC	PATC—AMTC (8)	4.0	4.0

Key: 1. Purpose of service

2. ATS - ATS 3. ATS - MTS

4. Route section of connecting lines

5. RATS - UVS (UIS)

6. UVS - UIS

7. RATS - UVS - UIS - RATS

8. RATS - AMTS 9. AMTS - RATS

10. Overall loss, dB

11. Along physical cable networks

12. Along channels of transmission systems

Table 1 shows normalized maximum values of the overall loss at a frequency of 0.8 kHz for SL, ZSL, and SLM.

In order to ensure stable transmission of interaction signals on frequencies of 0.7, 0.9, 1.1, 1.3, 1.5, and 1.7 kHz, there are additional requirements with respect to the overall loss for SL and ZSL (SLM) channels of the city telephone network which depend on the technical characteristics of code receivers.

A code receiver must normally work: through noncoil-loaded cable lines at a relatively lower level of the upper transmitted frequency by not over 4.3 dB when adjacent frequencies are transmitted, and by not over 10.4 dB in transmitting extreme frequencies used (0.7 and 1.7 kHz); through coil-loaded cable lines at a relatively low level of the lower transmitted frequency of not more than by 2.6 dB during the transmission of adjacent frequencies and not more than by 3.5 dB during the transmission of frequencies of 0.7 and 1.3 kHz.

Initial Data for Evaluating the Overall Loss of SL and ZSL Channels. In evaluating the overall loss of a section of a channel equipped with bridge-type amplifiers, their characteristic impedance  $\mathbf{Z}_{\mathbf{X}}$  on all frequencies is taken to be equal to the wave impedance  $\mathbf{Z}_{\mathbf{W}}$  of the cable circuit to which the amplifier is connected. Therefore, the changes in the operative attenuation of the amplifier-line joint are not taken into consideration in calculations. The value of the net loss of the section of the line with the amplifier is taken as the attenuation of the composite channel a.

The characteristic impedence of the section of the channel containing the KRR (KAMA) and IKM-30 equipment is established to be equal to 600 ohms. The image attenuation of the TCh channel of multichannel transmission systems is taken to be equal to its overall attenuation on all frequencies.

			Table 2					
<u></u>	۳. برآير	Параметры передвчя станционного четырех- полюсника (2)						
(1)	43010	12 <sub>x</sub> 1, 04 (3)	<sub>Фх</sub> . град (4)	а, дБ (5)				
	0.7 0.8 0.9 1.1 1.3 1.5	1243/957 1232/965 1198/975 990/960 600/907 675/850 543/790	ATCK +6.0/-14.15 +17.15/-15.25 +13.00/-17.00 +1.50/-17.25 -5.50/-12.50 -10.25/-7.50 -14.25/-3.00	0,60/0.65 0,56/0.62 0,55/0.59 0,53/0.55 0,55/0,52 0,58/0.50 0.61/0.48				
	0.7 0.8 0.9 1.1 1.3 1.5	923, 810/700 864/807/646 769/800/595 607/750/527 627/700/483 593/656/447	ATC-ДШ -50.00 -2.50/-58,50 -62.75/-4,00/-57,00 -62.30'-2.75/-55,50 -50.50'-4,50/-53,50 -59.50/-19.00/-32,00	1.44/0.77/0.41 1.36/0.74 0.44 1.31/0.75/0.46 1.23/0.77 0.45 1.21/0.50/0.52 1.21/0.50/0.52				

Note. First figure -- value for incoming connection, second figure -- for outgoing connection, third figure -- for tandem connection.

Key: 1. Frequency, kHz

- 2. Transmission parameters of a station quadripole
- 3. Ohms
- 4. Degrees
- 6. ATSK
- 5. Decibels
- 7. ATS DSh

The electrical characteristics of the channel elements, station quadripoles of ATS, and cable circuits are shown in Tables 2-3.

Fundamentals of the Method of Evaluating the Overall Loss of SL and ZSL Channels. SL and ZSL are composite channels whose connected elements are cable circuits, switching devices of ATS, UMT-type low-frequency amplifiers, and channels of transmission systems.

		Table 3	·						<del></del>
		1	(13)	Зизчение	параме	тра на ч	астоте,	кГц	
7.11	(1)	) — столначение параметра.  станица намерения	0,7	0.8	0.9	1.1	1,3	1.5	1.7
117-0,5	(2)	7. ALEKY (9) (10) 25. ON (11) 75. ON (12)	1.14 0.130 890 —44,7	0.140 830 44.63	1.26 0.148 785 44.58	1.40 0.162 715 —44.4	1,52 0,174 660 —44,25	1,63 0,165 620 —44,10	1,75 0,196 588 43,85
11'-0,7	(3)	2. дБ/км β. рад/км Z <sub>n</sub> . Оч Ф <sub>в</sub> . град	0.88 0.100 715 -44,10	0,95 0,110 672 44,00	1.00 0,114 630 —43,85	1,09 0,129 570 -43,55	1,14 0,134 535 43,28	1,19 0,141 500 43,00	1.23 0.148 480 —42,70
тпи-0,5	(4)	2. пБ/км р. галим Z в. Ом т <sub>в</sub> . град	1.11 0.128 975 44.55	1,17 0,135 9:7 44,50	1,24 0,142 865 44,40	1.38 0.158 775 44.28	1,50 0,174 700 —44,13	1,61 0,185 650 43,93	1,72 0,200 625 —43,72
11111-0.7	(5)	L км 1. разгим Z <sub>n</sub> . Ом q <sub>n</sub> . град	0,76 0,088 724 44,08	0.82 0.094 676 —43.93	0.87 0.100 639 —43.85	0,96 0,110 578 43,55	1,02 0,120 530 -43,30	1,09 0,130 500 -43,03	1,15 0,140 485 -42,75
13 9.5	(6)	и дБ км 1 рд. км Z <sub>b</sub> им Ф <sub>B</sub> прод	0.58 0.067 698 -43.8	553	616	0.73 0.084 557 —12.05	0.78 0.091 513 —41,20	0.83 0.103 460 -40.3	460
15-9,9	(7)	7. дБжи г раджи И <sub>в</sub> . Ом Ф <sub>в</sub> . град	0,51 0,058 641 43,0	600	0.57 0.066 566 542,5	0.63 0.073 512 -42,0	0.68 0.080 465 -41.5	430	0.74 0.095 410 0.40.5

Key: 1. Type of cable

2. TG-0.5

3. TG-0.7

4. TPP-0.5 5. TPP-0.7

6. TZ-0.8

7. TZ-0.9

8. Designation of parameter, unit of measurement

9. dB/km

10. rad/km

11. ohms

12. degree

13. Parameter value on frequency, kHz

The frequency dependence of the overall attenuation of SL and ZSL not containing coil-loaded cables is uniform within the frequency range 0.7-1.7 kHz. For lines formed by physical circuits, attenuation grows in proportion to  $\sqrt{f}$ lines containing UMT amplifiers or a transmission system, it practically does not depend on the frequency.

The increment of the channel attenuation due to mismatching of the input resistance depends little on the frequency in the frequency range of the interaction signals and has negative values. Therefore, the evaluation of the overall loss of SL and ZSL (SLM) composed of physical circuits of noncoil-loaded cables, UMT amplifiers and TCh channels formed by KPP equipment is done by simple summation of intrinsic losses on a frequency of 0.8 kHz, which ensures the fulfillment of the norms over the entire operating frequency range (0.7-1.7 kHz).

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In the case of composite channels containing coil-loaded cables, the increment of channel attenuation due to the mismatching of input resistance changes in a wavelike manner with frequency and can be positive. Due to this, the evaluation of the overall losses of such channels is done on a frequency of 0.8 kHz on which the attenuation of the speech channel is monitored, and on the operating frequencies of code receivers (0.7, 0.9, 1.1, 1.3, 1.5, 1.7 kHz).

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10,233

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LINE CIRCUIT EQUIPMENT OF THE K-1920P TRANSMISSION SYSTEM

Moscow ELEKTROSVYAZ' in Russian No 1, Nov 80 pp 6-11

[Article by N. Ye. Lugovskoy and G. N. Stepanov, submitted 1 Aug 1979]

[Text] The development and state tests of the domestic transistorized transmission system K-1920P have been completed. This equipment is intended for use in new main cable communication lines and for replacement of the operating tube equipment K-1920 for the purpose of raising technical and economic indexes of circuits and channels. The length of the repeater section remains the same -- 6 km. One of the main advantages of the new equipment is a longer distance (240 km) between attended repeater stations, which makes it possible to use manpower resources economically in operating the primary communication network. The K-1920P operates with the standardized terminal equipment with frequency division of channels OKOP and with terminal television equipment. The new transmission system satisfies the YeASS [Unified Automated Network of the Soviet Union] standards with respect to its basic parameters. Its series production has been started.

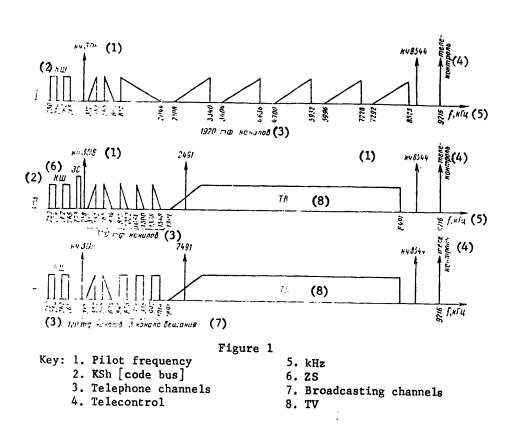
Composition and Purpose of the Equipment. The line circuit set K-1920P contains all types of equipment necessary for normal operation of main lines. This set includes: high-frequency equipment of the line circuit, including equipment for the line circuit of the distribution system K-24R and the remote power supply unit (DP); equipment for the link between operators (SS); telemechanics equipment (TM); equipment for the introduction, switching, and branching of television channels (TV); phase correction device, device for introducing a cable into NUP [unattended repeater station]; auxiliary equipment -- monitoring equipment, connecting lines with the television center, bay of DP filter for attended stations, device for checking DP circuits, and the telephone of the patrol lineman.

The K-1920P also includes the following: equipment for coupling the line circuit with the terminal conversion equipment OKOP [2] and television equipment [3]; equipment for isolating from the line frequency spectrum two lower in spectrum 60-channel groups of conversion of the video spectrum of television and sound signals into the line frequency spectrum.

High-frequency equipment of the line circuit is intended to work through coaxial cables with pairs of the normal diameter of 2.6/9.4 mm (chiefly through the KMB-4 cable). The use of the KMB-8/6 cable appears to be unlikely. However, if in some concrete case this proves to be practical, it is possible to use the K-300R working through 1.2/4.6 mm coaxial pairs.

Variants of the uses of the line spectrum of frequencies in the K-1920P transmission system are shown in Figure 1. The first variant corresponds to the organization of standard TCh [tone frequency] channels and group channels; the second variant shows the organization of 300 TCh channels, TV channels and sound channels (ZS); the third variant shows the organization of 120 TCh channels, a TV channel and three broadcasting channels one of which is intended for ZS.

It is evident that the second variant would be used chiefly in the reconstruction of old main lines when it is impossible to reduce the number of TCh channels, and the third variant will be used in the construction of new main lines for simplifying the organization of tandem operation by the line spectrum of TV and broadcasing channels when changing over to the K-3600 transmission system.



33
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In all variants of spectrum organization, its lower part is intended for operation of the equipment for noise power monitoring both in individual sections of OUP [attended repeater station]-OUP, and over the entire retransmission section (development of this equipment is not completed).

The K-1920P transmission system contains NUP [unattended repeater stations] of two types: basic and regulating. The basic NUP (their number on the main line constitutes approximately 75% of the total number of NUP) ompensate the attenuation of the repeater section of the cable 6+0.3 km long at a soil temperature of +8 degrees C and do not contain regulating devices except those compensating the variations in the lengths of the repeater sections.

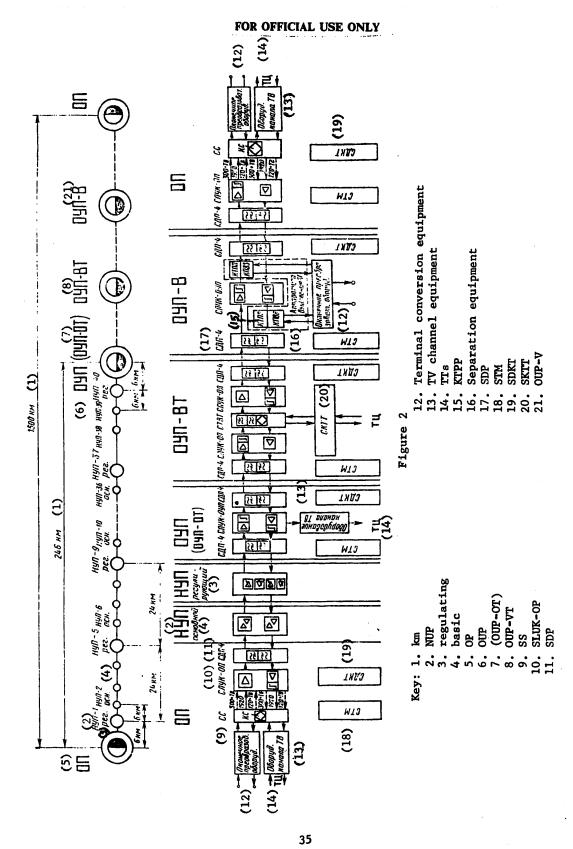
Regulating NUP are installed after each three basic NUP, preform their functions, as well as automatic gain control (ARU) depending on the changes in the attenuation of the cable at four repeater sections. ARU devices work when there are changes in the soil temperature and the level of the KCh [pilot frequency] of 8544 kHz. The regulation of gain by the soil temperature is ensured by preregulation at the output of the regulating NUP provided for in the transmission system.

The K-1920P transmission system uses OUP of the following types: without any additional functions (OUP); with TV channel branches (OUP-OT); with separation of groups of telephone channels (OUP-V); with separation and introduction of a TV channel (OUP-VT). The last type of OUP is used rarely and usually turns into a terminal point (OP).

ARU devices in OUP perform gain control only by the level of the KCh current. The devices of ARU by the basic KCh of 8544 kHz are supplemented by devices of ARU by an auxiliary frequency of 308 kHz. Moreover, it is possible to introduce a third KCh into the system, if it proves to be necessary from the results of tests and operation of the first main lines.

Remote monitoring of the condition of NUP amplifiers (their gain value) is performed from OUP through a special channel formed with the aid of filters included in each NUP between the routes of the opposite directions of transmissions. The filters have a passband of the order of 140 kHz and a medium frequency of 9216 kHz. Test signals, which are currents of 9216 kHz modulated by a quasi-random sequence of pulses, form in the transmitting part of the monitoring equipment. During the monitoring periods, they arrive from the OUP at the outgoing channel and return through the incoming channel. The time selection and analysis of the arriving response signals from all NUP of the maintenance section are performed in the receiving part of the monitoring equipment.

A block diagram of the main line with indications of maximum distances between individual points and the main composition of the equipment of stations of all types is shown in Figure 2, where SS, SLUK, SDP-4, STVT, STM, SKTT, SDKT are, respectively: coupling bay, line amplifier and corrector bay, remote power supply bay, television lead-in bay, telemechanics bay, TV channel correction bay, remote channel monitoring bay; TTs -- television center; KTPP -- direct communication channel set; KPVV -- tone frequency channel conversion, separation and lead-in set.



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The new K-1920P equipment complex is distinguished by a high degree of unification with equipment of the same type of the K-3600 transmission system [4]. The equipment of the distributive transmission system K-24R, remote power supply, service communication and telemechanics, device for cable entrance into NUP, and the majority of the devices of auxiliary equipment have been completely standardized. Containers of the high-frequency equipment of NUP and high-frequency equipment of OUP (OP) -- SLUK bays have been standardized structurally.

The equipment ensuring the organization of TV channels was developed with consideration of the experience in creating analogous devices of the K-3600 transmission system. The equipment for monitoring the operating conditions of NUP of K-1920P was developed on the same principles and structural basis as in K-3600, and differs only in the pilot frequencies, i.e., line frequency spectra.

Line Amplifiers. The equipment of the line circuit K-1920P for NUP uses two types of amplifiers: BUL (line amplifier unit) and BUR (regulating amplifier unit). The former is installed in the containers of the basic NUP, and the second in the containers of regulating NUP.

The line amplifier BUL has four stages and is fully transistorized: KT-335A -- in the first two stages, KT-610A -- in the third stage, and KT-904A -- in the fourth stage. The maximum undistorted power at the output of the amplifier is +16 dBn (300 mW).

The frequency characteristic of the amplifier is formed by line equalizer which is installed at the input of the amplifier and by the circuit connected in the feedback circuit. The amplitude-frequency characteristic (AChKh) of the amplifier corresponds to the attenuation of the coaxial cable with pairs of 2.6/9.4 mm 6 km long at to = +8 degrees C. The negative feedback circuit (OOS) of the amplifier has a manual gain control which makes it possible to change the gain depending on the length of the repeater section within the limits of 5.7-6.3 km in steps of 0.1 km. The input and the output of the amplifier has protective devices again the effects of lightning discharges. The power supply device provides for protection against the influence of LEP [electric power transmission lines].

The thermal noise level at the output of the amplifier adjusted to its input is 142 dBn on a frequency of 8.0 MHz, 137.5 dBn on a frequency of 4.0 MHz, and 126 dBn on a frequency of 0.5 MHz.

Nonlinearity attenuation at a fundamental frequency of 0.4 MHz and the output level on the second harmonic is not less than 88 dB, and on the third harmonic -- not less than 96 dB. Nonlinearity attenuation on the second harmonic at a fundamental frequency of 4.2 MHz is not less than 72 dB, and on the third harmonic at a fundamental frequency of 2.8 MHz -- not less than 91 dB. The attenuation of mismatching at the input and output of the amplifiers within the frequency range 2-8.6 MHz -- not worse than 25 dB, and within the range 0.3 -- 2.0 MHz -- not worse than 20 dB.

The BUR amplifier contains two amplifiers: linear and flat. The term "flat" is understood to be an amplifier with a frequency-independent gain characteristic. Moreover, BUR have: a receiver of the pilot channel of the fundamental KCh of 8544 kHz, a variable equalizer of ARU [automatic gain control] (VP ARU), and a device for gain control by the soil temperature (TRU).

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In the experimental model of NUP, the variable equilizer of ARU is included between the linear and flat amplifiers. Gain adjustment depending on the length spread of adjoining repeater sections (RRU) is done in the OOS circuit of the linear amplifier. In the series-produced equipment, the VP ARU circuit is included in the OOS circuit of the linear amplifier, and the RRU circuit is connected at the input of the BUR unit.

The TRU circuit is connected to the OOS circuit of the flat amplifier; its characteristics are controlled by a temperature sensor buried in the ground at a distance of approximately 10 m from the NUP.

The limits of gain adjustment on a frequency of 8.6 MHz by KCh are  $\pm 3.5$  dB; by the soil temperature --  $\pm 2.1$  dB. On the remaining frequencies of the range used, changes in the gain take place depending on  $\sqrt{f}$ .

Attenuation of nonlinearity of the BUR at the same output level as in the BUL, is normalized by 6 dB lower than in the BUL.

The OUP equipment has amplifiers of reception and flat amplifiers. With respect to its design and characteristics, the amplifier of reception is the same as the linear amplifier of BUR. The OOS circuit of flat amplifiers of OUP performs frequency-independent adjustment of the gain characteristic by  $\pm 1.0$  dB. Structurally, these variants of amplifiers are designed in application to the structure of the OUP bay.

Service traffic is organized through symmetrical pairs with conductors 0.9 mm in diameter. The number of the available conductors makes it possible to organize two four-wire channels through the KM-4 cable for station service traffic (PSS-1 and PSS-2) and one two-wire channel for section service traffic (USS). When the KMB-8/6 cable is used, the number of PSS channels increases to three. These channels ensure the transmission of frequency spectra from 0.3 to 3.4 kHz, and through the USS channel -- from 0.3 to 2.6 kHz.

Stations service traffic is organized in the OP-OP section of up to 1500 km between the OP themselves and between the OP and OUP; section service traffic is organized between OUP and NUP situated in sections adjoining the OUP through a coil-loaded pair with terminal amplifiers without intermediate amplifiers.

When the PSS is organized, intermediate amplifiers are installed every 30 km, as well as terminal amplifiers in OP and OUP. Selective calls through PSS channels are made with the aid of a TV [voice-frequency ringing] oscillator on 23 fixed frequencies; simultaneous calls to all OUP -- by a conference call on a frequency of 2460 Hz; a call to OUP from NUP -- with the aid of telemechanics devices; a call to NUP from OUP -- by a phonic signal with a frequency of 1020 Hz.

The service traffic equipment in OP and OUP is arranged on the SS bay. USS and PSS-2 channels can be used for communicating with NUP. A portable speaking set (PPU) containing terminal amplifiers of tonal frequency with gain control is connected to the SS channels in NUP. It is possible to connect to the USS channel through a socket situated on the TM [telemechanics] container or through a socket on the ground part of the NUP, and to the PSS-2 channel -- only through the socket on the container.

37

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Telemechanical Systems. The transmission systems K-3600 and K-1920P have two kinds of telemechanical systems; section telemechanical system (TMU) and main-line telemechanical system (TMM).

TMU systems are intended for receiving at the OUP (OP) announcement signals transmitted from NUP of adjoining service sections. These systems are used to transmit control commands to NUP in order to connect measuring oscillators (if this is required for monitoring the amplifiers of the NUP). The TMU equipment is located in telemechanics units (BTM) in NUP, and on separate telemechanics bays (STM) in OUP. The TMU operate through four signal conductors 0.9 mm in diameter.

OUP can receive eight announcement signals from each NUP, and NUP can receive six control commands from each OUP.

The TMU devices are using a time-cycle method of interrogating NUP in the service sections. For this purpose, the TMU equipment at OUP have an oscillator of timing pulses which are relayed to all BTM in NUP. OUP and NUP have signal distributors which operate synchronously and cophasally. Moreover, OUP have a distributor of NUP numbers. BTM in NUP have a remote supply system through phantom circuits organized through conductors set aside for the operation of TMU.

The main-line telemechanical devices TMM make it possible to monitor the condition of high-frequency channels on terminal or retransmission centers in all main line sections. Monitoring signals are formed in all OUP on the basis of signals arriving via the TMU system from NUP with addition of signals from local sensors.

The TMM system uses a reversible consecutive method of interrogating all OUP and OP of the main line, for which purpose there are regenerative repeaters of the amplitude and length of transmitted pulses every 18 km, at each third NUP.

As a result of the processing of all arriving signals, four announcement signals are transmitted to OP from each OUP; "normal", "warning", "damage", "emergency".

Two signal conductors  $0.9\ \mathrm{mm}$  in diameter are set aside for the operation of the TMM system.

Electric power supply for the K-1920P equipment, including remote supply, is received from a direct-current source of 24 V ±10%. The positive pole is grounded. The equipment of each transmission system is fed independently, and NUP equipment has a remote power supply with stabilized direct current through internal conductors of co-axial pairs with series connection of loads in the DP [remote power supply] circuit. The current in the DP circuit is 340 mA; the voltage delivered to the internal conductors of two coaxial pairs from attended stations, depending on the number of NUP, is from 50 to 1500 V.

The DP equipment is arranged on the remote power supply bay SDP-4.

Unattended Repeater Stations. The degree of automation of main coaxial cable lines is determined by the number of repeater stations operating without attending personnel. NUP equipment is located in underground metal tanks. As a result of this, insignificant fluctuations of the soil temperature at the depth of the location of the

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tank, in comparison with the fluctuations of the air temperature on the surface of the ground, do not affect the stability of the operation of the repeater equipment. This makes it possible to minimize the inspection of NUP by the operating personnel.

The fourth page of the magazine cover shows a NUP container K-1920P containing two units with line amplifiers for the basic NUP. Containers with NUP equipment are installed in tanks on special frames.

Small dimensions of transistorized equipment and smaller amounts of liberated heat than that of tube equipment made it possible to reduce the overall dimensions of the tank. The tanks developed by TsNIIS [Central Scientific Research Institute of Communications] have a diameter of 2.6 m and length of 2.4 m, and are intended for installation of containers of NUP of all types of the transmission systems K-3600 and K-1920P with the use of the cable KMB-4 or KMB-8/6, as well as the system VLT-1920.

Figure 3 shows the general appearance of a NUP installed in the ground. The NUP has five cable lead-in holes: two holes 60 mm in diameter for the main-line cable of both directions and three holes 20 mm in diameter for cables from ARU temperature sensors and a branch cable. Technicians enter the NUP through a manhole located in the upper part of the tank. Containers are fastened with bolts to special removable frames of the tank.

Two types of tanks are produced: with frames for fastening the equipment of K-3600 or K-1920P and with frames for fastening the equipment of VLT-1920. The tanks are covered on the outside with a special anticorrosive layer of bitumen with a fiber glass fabric. Moreover, in order to protect it against corrosion, negative potential is delivered on the body of the tank with the aid of four magnesium electrodes buried near the NUP.

The sensors located in the NUP transmit to the OUP with the aid of telemechanical devices signals about the opening of the manhole, about the appearance of water in the NUP, and about the dropping of air pressure in the cable, and from the points with ARU -- signals about the lowering of the KCh level, etc.

Main cables entering the tank terminate in terminal cable devices (UOK) to which containers are connected with the aid of flexible station cables.

Arrangement of Equipment. The equipment of OP and OUP is arranged in bays of a functional unit design of 2600 X 680 X 225 mm. The external appearance of the bay of line repeaters and correctors of terminal points (SLUK-OP) is shown on the fourth page of the magazine cover.

Results of Line Tests. The state tests of the specimens of the line equipment complex of K-1920P conducted in an experimental section of the main line of 115.7 km containing the equipment of 14 basic and 5 regulating NUP, as well as the equipment of OP and OUP showed the following.

The divergence between the AChKh of the gain of line repeaters and the attenuation of the cable sections is not over 0.1 dB, the power of total interference of the line circuit in the band of the TCh channel at the loading of the channels by white noise signals of 50 microvolts at the point of the relative zero level is not more than 1 pkW·psoph/km.

39

The equipment ensures high stability of the overall loss of the line circuit. The dynamic characteristics of the line circuit correspond to the MKKTT [International Telegraph and Telephone Consultative Committee] standards. Protective devices ensure normal operation of the line with all types of regulated external influences.

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10,233

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## PUBLICATIONS, INCLUDING COLLECTIONS OF ABSTRACTS

ABSTRACTS FROM THE COLLECTION 'MODERN METHODS OF SIGNAL PROCESSING'

Moscow TRUDY MOSKOVSKOGO ORDENA LENINA ENERGETICHESKOGO INSTITUTA: TEMATICHESKIY SBORNIK: SOVREMENNYYE METODY OBRABOTKI SIGNALOV in Russian No 455, 1980 (signed to press 26 May 80) pp 2, 147-154

[Annotation and abstracts of articles from the collection "Transactions of the Moscow Order of Lenin Power Engineering Institute: Topical Collection: Modern Methods of Signal Processing", edited by Professor G. D. Lobov, 300 copies, 154 pages]

[Text] The material of this collection treats urgent problems of modern science and technology. Investigation of new principles of signal processing and their practical application are developing widely. This is connected with the development of new functional assemblies for radio electronics. This collection presents the results of studies of some general principles of signal processing, functional processing devices with the use of new physical principles and the results of work on the synthesis of new materials for such devices. In accordance with these directions of studies, the material of the collection is divided into three topical sections.

The articles in this collection are the results of original studies by the authors on the above-mentioned problems. On the whole, the collection will be useful for a broad section of scientists and engineers working in the area of radio electronics and related areas of science and technology.

UDC 621.371

ON THE INFLUENCE OF A DISPERSIVE MEDIUM IN A RADIO COMMUNICATION CHANNEL ON MATCHED RECEPTION OF PHASE-MANIPULATED SIGNALS

[Abstract of article by Kramm, M. N.]

[Text] The author examines the passage of radio signals through a section of a radio communication channel which includes a dispersive medium and an optical processing device -- matched filter. The results of calculations are given for signals with phase manipulation by the Barker code. The shape of the signal envelope at the output of the matched filter is analyzed for various combinations of the parameters of the signal and the medium.

41

UDC 621.391.81

APPROXIMATION OF THE SHAPE OF ANOMALOUS OVERSHOOTS OF THE PHASE, FREQUENCY, AND THE ENVELOPE OF THE SUM OF THE SIGNAL AND NOISE BY THE MOST PROBABLE TRAJECTORIES

[Abstract of article by Zhukov, V. P., Ivanova, N. N., and Razumov, L. A.]

[Text] The authors found relations describing changes in the phase, frequency, and the envelope of the sum of the signal and narrow-band Gaussian noise at phase jumps by  $\pm 2~\%$ . The solution is done by the method of the most probable trajectories. Distribution functions of random parameters determining the shapes of such trajectories are given.

UDC 621.391

ACCURACY OF THE IDENTIFICATION OF THE PARAMETERS OF A MULTIBEAM COMMUNICATION CHANNEL AGAINST THE BACKGROUND OF QUASI-STATIONARY GAUSSIAN INTERFERENCE

[Abstract of article by Arapov, S. M.]

[Text] The author examines the block diagram of subsystems of the identification of the parameters of a multibeam receiver. Basic relations describing the operation principle are obtained. Conditions are found in which the accuracy of identification increases. The dependence of the accuracy of identification on the signal phase is found.

UDC 621.398.1

INVESTIGATION OF SIGNAL PROCESSING ERRORS IN MEASURING SHORT DISTANCES WITH THE AID OF ULTRASOUND

[Abstract of article by Ragozin, Yu. D., Volosskiy, V. P., and Pakhomov, V. P.]

[Text] The authors examine basic sources of errors occurring in measuring short distances with the aid of ultrasound. They show that measurement errors under certain conditions can be reduced to decimal fractions of a millimeter.

UDC 621.391.272:539.143

METHOD ERRORS IN PROCESSING SIGNALS WITH THE AID OF SPIN ECHO

[Abstract of article by Kalinin, V. A.]

[Text] The author examines errors in the realization of the signal correlation and delay functions connected with the properties of pulses of primary and stimulated echo. Errors are found in solving Bloch equations by the method of successive approximations. The author determined the errors connected with an insufficient amplitude of the reading pulse and excessively great intensity of the input signals. Requirements for the parameters of signals and the reading pulse are formulated.

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UDC 620.179.162

ANALYSIS OF THE SPECTRUM OF PULSED SIGNALS DURING ULTRASONIC CHECKING OF MULTILAYER MEDIA

[Abstract of article by Skogorev, M. Ye.]

[Text] The author examines the spectrum of a signal occurring during ultrasonic checking of layered media. It is shown that the spectrum can be used for obtaining information about the gluing layer. The bifurcation of the resonance peaks of the spectrum revealed in the experiment is connected with the resonances within the layered material itself.

UDC 621.38

THERMAL STABILITY OF MEMORY DEVICES USING CHALCOGENIC GLASS-LIKE SEMICONDUCTORS

[Abstract of article by Borovov, G. I. and Voronkov, E. N.]

[Text] The authors give the results of experimental studies of a memory matrix K524RPI. It is shown that thermal properties of an instrument are determined chiefly by the properties of the glass. They obtained the parameters of a kinetic equation of crystallization which make it possible to evaluate the service life of an instrument at higher temperatures. The results indicate the necessity of using glass with high values of activation energy.

UDC 621.379.54

COMPUTATION OF FILTERS ON ACOUSTIC SURFACE WAVES WITH NONEQUIDISTANT COUNTER-PIN CONVERTERS

[Abstract of article by Kartashev, V. G.]

[Text] The author examines the synthesis method of band filters on acoustic surface waves with two identical nonequidistant counter-pin converters. He concludes that, in order to obtain a prescribed frequency characteristic close to rectangular, it is necessary to achieve the amplitude apodization of at least one converter. The characteristics of filters synthesized by this method are given.

UDC 621.37:534

TEMPERATURE DEPENDENCE OF THE PHASE AND AMPLITUDE OF THE OUTPUT SIGNAL IN AN AMPLIFIER OF ACOUSTIC SURFACE WAVES

[Abstract of article by Gavrilin, V. I., Gulyaev, A. M., Karpeyev, D. V., and Bash-kirov, A. M.]

[Text] The authors give the results of their experimental studies on the characteristics of an amplifier of acoustic surface waves in a range from 100 to 400 K. It is shown that, when the temperature decreases, the losses increase and there are changes in the phase of the signal at the output of the amplifier. It is concluded that the increase of losses is connected with the increase of acoustoelectronic interaction.

43

UDC 620.179.1

PROBLEMS OF PROCESSING ROENTGEN-TELEVISION SIGNALS

[Abstract of article by Volkov, A. V., and Zorin, A. Yu.]

[Text] This article gives the results of studies on the use of series-produced ultrasonic delay lines in television roentgen defectoscopy of reinforced objects. Circuit diagrams of matching and use in the radio-pulse and video-pulse modes are given. Requirements for the delay lines for the above method of defectoscopy are formulated. It is indicated that it is necessary to develop delay lines for a carrier frequency of about 30 MHz.

UDC 621.391.272:539.143

EFFECTS OF THE SHAPE OF THE WORKING BODY ON THE TRANSMISSION COEFFICIENT OF A SPIN FUNCTIONAL ASSEMBLY

[Abstract of article by Ivanov, Yu. V., Lobov, G. D., and Shtykov, V. V.]

[Text] On the basis of the known solution of the Bloch equation, the echo signal voltage was found at the output of a spin functional assembly with consideration for the configuration of the working body and the registering coil. Results of computer computations for a toroidal core are given. It is shown that cores with a minimum radius-thickness ratio should be used in order to increase the transmission coefficient.

UDC 621.179.1

ELECTROACOUSTIC CONVERTERS OF SYSTEMS FOR OPTIMAL PROCESSING OF ULTRASONIC SIGNALS

[Abstract of article by Aksenov, V. P., and Popko, V. P.]

[Text] On the basis of the analysis of methods of optimal processing of signals, the authors determined the requirements for the parameters of electroacoustic converters for instruments of ultrasonic nondestructive inspection. Designs of proposed modifications of wide-band electroacoustic converters are described and experimental data confirming the correctness of the initial theoretical premises are given.

UDC 621.37:534

ON THE POSSIBILITY OF USING THE ACOUSTOMAGNETOELECTRIC EFFECT FOR STUDYING THE PROPERTIES OF AMPLIFIERS OF ACOUSTIC SURFACE WAVES

[Abstract of article by Gavrilin, V. I., Gulyayev, A. M., Karpeyev, D. V., and Praporshchikov, V. V.]

[Text] The authors propose a new method for studying the characteristics of amplifiers of acoustic surface waves. Results of their studies are given. It is shown that this method will be useful for studying the parameters of carriers in the area of acoustoelectronic interaction not only of amplifiers, but also of other instruments with a layered piezoelectric-semiconductor structure.

44

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UDC 621.373.12

SELF-OSCILLATOR ON THE DELAY LINE OF ACOUSTIC SURFACE WAVES WITH EXTERNAL SYNCHRONIZATION

[Abstract of article by Dvornikov, A. A., Ogurtsov, V. I., and Utkin, G. M.]

[Text] The authors examine a self-oscillator on a delay line of acoustic surface waves with external synchronization. They analyze shortened equations of such an oscillator. For the cubic approximation of the dependence of the output current of the active element, characteristics of a synchronized oscillator of this type are given. Qualitative experimental confirmation of the theoretical results is obtained.

UDC 621.372.8

FILTRATION OF PARASITIC WAVES IN MEANDER-TYPE STAGGERED PRINTED DELAYING SYSTEMS

[Abstract of article by Berezin, V. M., Guttsayt, E. M., Mal'tov, V. N., and Skripov, A. A.]

[Text] The authors investigated the possibility of effective suppression of parasitic waves of antiphase oscillations in a two-storeyed meander by means of filters on the basis of subsidiary delaying systems of printed ring vibrators which in combination with a clamp-type systems make it possible to obtain losses of 1.3-1.5 dB per cell, while the losses at the working type of oscillations are not over 0.05 dB per cell.

UDC 621.38

DEVICE FOR AUTOMATIC REGISTRATION OF THE DRIFT OF VOLT-FARAD CHARACTERISTICS IN THE PROCESS OF TEMPERATURE AND FIELD TESTS

[Abstract of article by Sobolev, M. V., Korchagin, S. N., Kolomeychuk, A. S., Varlashov, I. B., and Soldatov, V. S.]

[Text] The authors give a block diagram of a device for registering volt-farad characteristics in the process of temperature and field tests. They describe the method of measurements. The developed device makes it possible to register a shift in the volt-farad characteristics of 0.05 V at a time constant of not less than 0.1 sec. Comparison with the standard method indicated the correctness of results obtained by this method.

UDC 621.373.826

OPERATIONAL CHARACTERISTICS OF A PYROMAGNETIC RECEIVER OF INFRARED RADIATION

[Abstract of article by Lobov, G. D., Shtykov, V. V., and Drugov, L. V.]

[Text] Results of experimental studies of a pyromagnetic detector are given. The functional scheme of the experiment is described. Experimental relations are compared with theoretical relations obtained for the simplest model of this phenomenon. Recommendations are given for improving the characteristics of the receiver.

45

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UDC 534.422.8.063

PROBLEMS OF THE METHODS OF ROENTGENOSPECTRAL MICROANALYSIS OF MULTICOMPONENT GARNET SYSTEMS

[Abstract of article by Tsvetkova, A. A., Yegorov, S. K., and Kon'kova, M. M.]

[Text] The authors examine the peculiarities of conducting quantitative analysis in the system  $Gd_3$ -x  $Ca_xGa_3$ -y $Zr_yO_{12}$ . Single crystals and polycrystals were used as samples. Problems of the methods of preparing samples and standards are discussed. Stoichiometric peculiarities of this system are examined.

UDC 535.212.

DETERMINATION OF THE SPEED OF THE MOTION OF TRANSLATION OF CYLINDRICAL MAGNETIC DOMAINS BY THE PHOTOELECTRIC METHOD

[Abstract of article by Nikolayev, L. V.]

[Text] The author proposes a method making it possible to determine the speed of motion of cylindrical magnetic domains in a Bi-containing film of ferrite garnets without possible errors connected with the inertia effect during the motion of translation of the domain (ballistic effect). A scheme for the realization of the proposed method is given.

UDC 621.365.6

EFFECTS OF TECHNOLOGICAL CONDITIONS ON THE CONFIGURATION OF THE THERMAL FIELD IN MELTS DURING CZOCHRALSKI CRYSTAL GROWTH FROM A COLD CRUCIBLE

[Abstract of article by Balbashov, A. M., and Zavartsev, Yu. D.]

[Text] Negative results during the growing of crystals from a cold crucible are caused by unfavorable distribution of temperature in the melt which is characterized by the overheating of the central areas. The authors show that it is possible to obtain a favorable configuration of the thermal field in the melt by controlling the size and direction of the convection flow.

UDC 621.317

STATIC MAGNETIC CHARACTERISTICS OF Bi-CONTAINING GARNET FILMS

[Abstract of article by Pavlova, S. G., and Cherkasov, A. P.]

[Text] The authors give the results of their experiments in the investigation of static magnetic properties of garnet films of the systems (Yb Gd Bi)3 (Fe Al)5  $0_{12}$ ; (YBi)3 (Fe Ga)5  $0_{12}$ . They studied the behavior of a domain structure in the temperature interval from -100 degrees C to the Curie point. Conclusions are made on the optimization of the compounds in the studied systems in order to increase the thermal stability of magnetic characteristics.

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UDC 621.317:535.212

SYNTHESIS OF MULTILAYER FILMS IN THE SYSTEM (YBi)<sub>3</sub> (FeCa)<sub>5</sub> O<sub>12</sub> AND INVESTIGATION OF THE BLIND DOMAIN STRUCTURE IN THEM

[Abstract of article by Cherkasov, A. P., Chervonenkis, A. Ya., Cheparin, V. P., and Shupegin, M. L.]

[Text] The authors describe methods of synthesizing multilayer Bi-containing garnet films possessing a prescribed profile of magnetic constants over their thickness. They give results of their studies on variants of blind domain structures in such films. They suggest the uses of blind domains in the information and measuring techniques.

UDC 621.317:535.212

OPTICAL ABSORPTION SPECTRA OF EPITAXIAL FILMS OF Bi-CONTAINING GARNETS

[Abstract of article by Balbashov, A. M., Bakhteuzov, V. Ye., Tsvetkova, A. A., and Chervonenkis, A. Ya.]

[Text] The authors analyzed absorption spectra of epitaxial films of Bi-containing garnets in the 0.4-2.5 micron range. They revealed individual contributions into the absorption of ions Bi, Pb<sup>2+</sup>, Pb<sup>4+</sup>, Fe<sup>2+</sup>, Fe<sup>4+</sup>, Co<sup>2+</sup>. Corresponding curves of additional absorption make it possible to evaluate the concentration of the above ions and optimize the conditions of synthesis.

UDC 621.317:533.9.74

SELECTION OF THE PARAMETERS OF A SIMICONDUCTOR FOR OBTAINING THE MAXIMUM MAGNETIC MOMENT OF THE ELECTRONS AT A ZERO MAGNETIC FIELD

[Abstract of article by Gratsianskaya, Ye. I.]

[Text] The author calculated the magnetic moment of electrons of the conductivity zone of a semiconductor in the absence of a constant magnetic field when an electromagnetic wave with circular polarization falls on the specimen.

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CSO: 1860/211

ELECTRICAL ENGINEERING HANDBOOK (IN THREE VOLUMES): VOLUME 1; GENERAL ASPECTS. ELECTRICAL ENGINEERING DATA

Moscow ELEKTROTECKHNICHESKIY SPRAVOCHNIK (V TREKH TOMAKH): TOM 1; OBSHCHIYE VOPROSY. ELEKTROTEKHNICHESKIYE MATERIALY in Russian 1980 (signed to press 5 Jun 80) pp 2, 5-7, 23-24, 69, 106-7, 175, 222, 248-9, 263, 287, 316, 325, 351, 397, 422, 481

[Annotation and section tables of contents from book "Electrical Engineering Hand-book (in Three Volumes): Volume 1; General Aspects. Electrical Engineering Data", edited by I. V. Antik, Izdatel'stvo "Energiya", 80,000 copies, 520 pages]

[Text] This volume contains information on the theoretical bases of electrical engineering, on the electrophysical properties of conductors, semiconductors and insulators, on the general aspects of electrical equipment and safety engineering, on relevant papers on electrical engineering, on the measurement of electrical and magnetic quantities and on cable equipment. Instructions are given regarding the execution of drawings and circuit designs. The preceding fifth edition of this handbook was published in 1974.

It is intended for a broad circle of electrical engineers.

Contents			Page
Foreword to the Sixth Edition of the Electrical Engineering Handbook	•	•	-3
Section 1. Units of Physical Quantities. Most Important Physical Constants	• ,	•	7
Section 2. Basic Rules for Drawing Up Design Documentation	•	•	23
Section 3. Basic Information on Electrophysics	•	•	69
Section 4. Theoretical Bases of Electrical Engineering	•	•	106
Section 5. Measurement of Electrical and Magnetic Quantities	•	•	175
Section 6. General Aspects of Electrical Equipment	•	•	222
Section 7. Ascertaining the Effectiveness of the New Technology .	•	•	248

## 48

	Section	8.	Safety of Electrical Installations and Equipment	•	•	•	•	263
	Section	9.	Electrical-Insulation Materials	•	•		•	287
-	Section	10.	Semiconducting Materials		•		•	316
	Section	11.	Magnetic Materials	•			•	325
	Section	12.	Conducting Materials, Noninsulated Conductors and	Buse	28			351
	Section	13.	Electrical-Insulation Designs and Insulators					397
	Section	14.	Cable Equipment					422
	Section	15.	Semiconductor Instruments and Microcircuitry	•	•			481
	Subject	Inde	×					512
•			Volume Two Contents					
	Electric	al E	quipment					
•	Section	16.	Resistors, Capacitors, Reactors					
	Section	17.	Transformers and Autotransformers					
	Section	18.	General Aspects of Electrical Machinery					
	Section	19.	Alternating-Current Electrical Equipment					
	Section	20.	Direct-Current Electrical Equipment					
	Section	21.	Electrical Machinery in Automatic Equipment					
	Section	22.	High-Voltage Switching and Protective Apparatus					
	Section	23.	Current and Voltage Transformers					
	Section :	24.	Low-Voltage Apparatus					
	Section 2	25.	Unitized High-Voltage Installations					
	Section 2	26.	Rectifier Transducers of Electric Power					
	Section 2	27.	Chemical Current Sources and Their Application					
			Volume Three Contents					
	Book 1.							
	Production	on, I	Transmission and Distribution of Electric Power					•

#### FOR OFFICIAL USE ONLY

- Section 28. General Aspects of Power Systems
- Section 29. Circuit Designs for Electrical Connections of Electric Power Stations and Substations
- Section 30. Designs for Switching Equipment
- Section 31. High-Voltage Electric Networks
- Section 32. Transmission of High-Voltage Alternating and Direct Current
- Section 33. Designs of Overhead and Cable Lines
- Section 34. Electric-Power Supply to Cities and Rural Areas
- Section 35. Transients in Electric-Power Systems
- Section 36. Short-Circuit Currents and the Selection of Apparatus
- Section 37. Overvoltages and Overvoltage Protection
- Section 38. Automation in Power Systems
- Section 39. Relay Protection
- Section 40. Control, Monitoring and Alarm Systems at Power Stations and Substations
- Section 41. Technical-Economic Indicators in Power Systems

#### Book 2.

- Utilization of Electric Power
- Section 42. Electric Drive
- Section 43. Automatic Control of Electric Drives
- Section 44. Electric Drive of Commonly Used Industrial Mechanisms
- Section 45. Electrothermal Equipment
- Section 46. Electrical Engineering Equipment
- Section 47. Electric Welding Equipment
- Section 48. Electric Lighting
- Section 49. Electrical Equipment in Explosion-Hazard Industries
- Section 50. Electric-Power Supply to Industrial Enterprises

#### JAL USE ONLY

Page

8

10

10 10

11

12

13

14 .

14

15

15

15

16 16

18

18

18

20

Section	51.	Electric Transport					
Section	52.	Electrical Equipment for Automobiles					
Section	53.	Electric Appliances					
Section	54.	Electric Hand and Portable Machines (Electric T	'oo1	.s)			
Section	55.	Industrial Radio Interference					
Section	56.	Industrial Noise					
		Volume One Contents					
		volume one contents					
Section	1.	Units of Physical Quantities. The Most Importan	t P	hys	ica	<b>a</b> 1	
1.1.	Inte	rnational system of units (SI)	•	•	•	•	•
	Basic	a CT .m.i.s.					•
	Addi	tional units	•	•	•	•	•
1.2.	Unita	s of mechanical quantities			•	•	•
	Syste	om of imilia da alles i .		:	•	•	•
	Units	s of mechanical quantities in the International	Sys	tem	· of	:	•
	Units	s outside the system accepted for use on a level	wi	th	SI	•	•
		4 4 -					

Units outside the system accepted for use in special areas

List of units of the moment of inertia and the flywheel moment . Metric and nonmetric units of length, area, volume, mass, work and energy, power, pressure, velocity, frequency of rotation .

1.4. Units of thermal and light quantities characterizing ionizing

Units of quantities characterizing ionizing radiation and radioactivity in the International System of Units . .

Units of electrical and magnetic quantities . . .

Units of thermal quantities in the International System of Units.

Units of light quantities in the International System of Units .

Factors and additive terms for forming multiple and fractional 

1.7. Most important physical contants (fundamental constants) . .

1.8. D. I. Mendeleyev's periodic chart of the elements . . . .

Units of electrical and magnetic quantities accepted on a level 

1.9.	Letter designations of the major electrical and magnetic quan-		,	20
	tities	•	/e	22
Bibliog	raphy	•	•	23
Section	2. Basic Rules for Drawing Up Design Documentation			
2.1.	General information and basic points of the Unified System of			24
	Design Documentation	•	•	26
2.2.	Basic requirements for producing drawings	•	•	
	Formats	•	•	26
	Scales	•	•	27
	Lines	•	•	27
	Basic legends	•	•	27
	Lettering	•	•	27
	Depiction of objects	•	•	27
	Laying out dimensions and legends	•	•	29
2.3.	Producing drawings at the design stage	•	•	30
	Stages of development		•	30
	Detail design		•	31
	Working documentation		•	31
2.4.	General rules for executing design documentation of various item	18		33
	Rules for executing design documentation for objects with elec-			
	trical wiring		•	33
	Rules for executing drawings for electrical wiring			33
	Rules for executing drawings of bundled conductors, cables and			
	individual conductors			34
	Drawings of objects with electrical windings and magnetic			
	circuits			35
	Depiction of printed circuit boards			36
2.5.	Rules for graphic layout of electric circuits			37
2.5.	Circuit classification			37
	General rules for executing circuits			37
	Rules for executing block diagrams and functional diagrams .	•		38
	Rules for executing schematic diagrams			38
	Rules for executing diagrams of connections and switches	•		41
	Rules for executing diagrams of connections and switches	•	•	42
	Rules for executing general and layout diagrams	•	•	43
2.6.	Graphic symbols used in circuit designs	•	•	43
	Commonly used symbols (according to State Standard 2.721-74).	•	•	43
	Electric machines (according to State Standard 2.722-68)	•	•	43
	Inductance coils, reactors, chokes, transformers, autotrans-			
	formers and magnetic amplifiers (according to State Standard			45
	2.723-68)	•		
	Current collectors (according to State Standard 2.726-68)		•	47
	Arresters. Fuses (according to State Standard 2.727-68)	•	•	47
	Resistors. Capacitors (according to State Standard 2.728-74).	•	•	48
	Electrical measuring instruments (according to State Standard			
	2.729-68)	•	•	51
	Semiconductor devices (according to State Standard 2.730-73) .	•	•	51
	Light sources (according to State Standard 2.732-68)	•	•	54
	Chemical current sources (according to State Standard 2.742-68)	•	•	55
	Electrothermal equipment and installations (according to State			
	Standard 2.745-68)			55

## FOR OFFICIAL USE ONLY

	shapes of pulses (according to State Standard 2.750-68).  Lines of electrical communication, conductors, cables and buses (according to State Standard 2.751-73).  Switching devices and contact connections (aacording to State Standard 2.755-74).  Sensing portion of electromechanical devices (according to State Standard 2.756-76).  Binary logic elements (according to State Standard 2.743-72).  Analog and analog-digital computers (according to State Standard 23335-78).	5 5 6 6
Section	3. Basic Information on Electrophysics	
3.1.	Plantramamatic phanamana in an I am I am I am I	7(
	Band theory	70
	Intrinsic and extrinsic conduction in semiconductors.	7:
	Contact phenomena at the metal-semiconductor boundary	7:
	Electron-hole movement	74
	Galvanomagnetic phenomena.	7:
	Magneto-optical phenomena.	7
	Internal photoeffect	7
	Luminescence	78
3.2.	Kleetrie processes in inculations	79
	Polarization of insulators	79
	Insulator conductivity	8:
	Insulator breakdown.	82
3.3.	Electromagnetic phenomena in ferromagnetic substances	8.5
	Magnetization processes	8.5
	Eddy currents during magnetic reversal and magnetic viscosity.	86
3.4.	Superconductivity	87
	Basic information	87
	Electromagnetic properties of superconductors	88
	Areas of application for superconductors	90
3.5.	Electron emission	90
	Thermoionic emission	90
	Autoelectronic emission	91
	Photoemission	91
	Secondary emission under the influence of positive ions, metastable	71
	atoms and discharge radiation	92
3.6.	Movement of electrons and ions in a vacuum and in cases	93
	Electric current in a vacuum.	93
	Collisions of electrons and ions with atoms and molecules of a cas	93
	Movement of electrons and ions in a gas	94
3.7.	Types of electric discharge in gases	96
	Dark discharge and evolution of a self-maintained discharge	96
	Glow discharge	97
	Arc discharge	97
	Corona and spark discharges	98
	Gas plasma	99
3.0.	rundamentals of quantum electronics	101
3.9.	Electromagnetic phenomena in a moving conducting medium	103

53

		103
	Magnetohydrodynamic equations	104
	Magnetic pressure, electromagnetic pumps and accelerators	104
	Magnetic pressure, electromagnetic pumps and accelerators	
Bibliog:	raphy	
•	The state of the s	
Section	4. Theoretical Bases of Electrical Engineering Linear circuits for direct and sinusoidal currents	107
4.1.	Linear circuits for direct and sinusoidal currents	107
	Direct current	
	Instantaneous, root-mean-square and mean values of sinusoidal	107
		108
	Representation of sinusoidal quantities through complex numbers .	108
	Parameters of circuit element and equivalent circuits	109
	Passive two-terminal networks	110
	Resistance and conductance of two-terminal networks	113
	Active two-terminal networks	114
	Mutual inductance	114
	Power	115
	Electric current diagrams and their flow sheets	116
	Kirchoff's law	117
	Topographical vector diagrams	118
	Woltage and current resonance	119
	Application of Kirchoff's laws for calculating electric circuits .	
	Toon-ourrent method	120
	Nodal-notential method	121
	Application of tonographical methods of calculation · · · · · ·	123
	Mittagle of entry length exchange of sources of current and e.m.I	124
	Active two-terminal network method	124
	Morthad of gunernosition	124
	Transforming circuit element connections with deltas and stars	125
	Cimilitude method	
	Principle of compensation	126
	Theorem of variations	126
	Peciprocity theorem	126
	Times relationships	126
	Transfer functions	126
	Designating voltages on circuit diagrams	126
	Cdomed_flow graphs	127
4.2.		129
4.2.	minus shows greatom with a neutral wire	129
	Three-phase system with a neutral wire	130
	Method of symmetric components	133
, ,		131
4.3.	Expansion of periodic functions into trigonometric series	133
	Root-mean-square and mean values of nonsinusoidal periodic quan-	
	titles	133
	Nonsinusoidal factors	
	Nonsinusoidal ractors	134
	Equivalent sinusoids	134
	Higher harmonies in three-phase circuits	
	Intercoupling of magnetic flux. E. m.f. of induction, voltage and	13.
	aurrent in a coil with a steel magnetic circuit	

54

## IAL USE ONLY

	Equivalent circuit of a coil with a steel magnetic circuit when	
	losses are considered	. 13
4.4.	Transients	. 13/
	Classic method	. 130
	Classic method .  Transients in the simplest circuits	. 13
	method of state variables.	1/4
	Operational method	. 149
	neaviside's formulas	. 14
	Dunamel integral.	1/.
	rourier integral.	140
	Cutting in a network branch	1/40
	Cutting off a network branch.	140
4.5.	Passive four-terminal networks, filters and lines	. 140
	Nonsymmetrical four-terminal networks	1/40
	Symmetrical four-terminal networks	152
	rilters	154
	Ladder circuits and other connections for four-terminal networks	. 159
	Line with losses	. 157
	Determining line constants by means of no-load and short-circuit	
	tests	157
	Simplified formulas	. 158
	Low-loss line	. 158
	NO-loss line	159
	Transients in circuits with distributed constants.	. 150
	Reflection of a square wave from the end of a line	. 160
	Incidence of square wave at the junction of two lines	. 161
4.6.	Electric field	. 163
	Coulomb's law	163
	Field strength. Potential	. 163
	rolarization, shift	. 165
	Current, current density	. 166
	Laplacian and adiabatic equations. Boundary conditions	166
	Distribution of potential in a system of charged conducting bodies	. 166
	Capacitance	. 167
	Direct capacitance	. 167
	Modeling	. 167
	Electrostatic field energy. Forces	. 167
4.7.	Magnetic field	. 167
	Magnetic induction and magnetic flux	167
	Magnetization. Magnetic field strength	. 168
	Ampere's circuital law	168
	Magnetic field potentials. Boundary conditions	. 168
	Energy. Electrodynamic forces	. 169
	Electromagnetic induction (e.m.f. induction)	160
, ,	Self-induction and mutual induction	. 169
4.8.	Electromagnetic field	. 169
	Maxwell's equations for an electromagnetic field in a stationary	
	medium	. 169
	medium	. 170
	POTABLIC	

55

	Maxwell's equations in combined form			•	•	170
	Assessing losses. Umov-Pointing theorem	,				170
	Resistance, capacitance, inductance			_		170
4.9.	Resistance, capacitance, inductance		• •			170
	Resistance of a conductor.	'	• •	•	•	170
	Capacitance of capacitors and conductors	,	• •	•	•	
	Electrostatic analog method	•	• •	•	•	171
	To tour many of god to			• .		1/4
	Capacitance, inductance and mutual inductance in overload	ad	lines			173
	Coaxial cable inductance					174
	Coaxial cable inductance		•			174
Bibliogr	aphy	•	• •	•	•	-17
Section	5. Measurement of Electrical and Magnetic Quantities					
5.1.	Basic concepts	•		•	•	175
5.2.	Types and methods of measurement	•		•	•	176
5.3.	Means for measuring electrical quantities and their bas	ic	chara	te	r-	
3.3.	istics					177
- •	1STICS	•	•	•		182
5.4.	Measurement of voltages and currents	•	• •	•	•	183
5.5.	Measurement of electrical circuit parameters	•		•	•	
5.6.	Measurement of power	•	• •	•	•	187
5.7.	Measurement of power	• ່		•	•	190
5.8.	Direct and alternating current bridges		• •		•	190
5.9.	Direct and alternating current potentiometers					191
	Direct and alternating current digital voltmeters.	_		_		191
5.10.	Direct and afternating current digital voicemeters.	•	•	•	-	193
5.11.	Integrated digital instruments	•	• • •	•	•	194
5.12.	Panel-mounted analog instruments	•	• •,	•	•	
	General information	•	• •	•	•	194
	Small-size instruments	•		•	•	197
	large-size instruments			•	•	198
5.13.	Measurement errors. Processing of direct measurements					201
5.15.	Methods of describing random errors					202
	methods of describing random errors	•		-	-	203
	Confidence intervals	•	•	•	•	204
	Processing direct measurements	•		•	•	
	Systematic errors in measurement	•	• •	•	•	
5.14.	Representation of measurement results	•		•	•	207
5 15	List of State Standards for electrical measurement inst	:ru	ments	•	•	200
5 16	Methods of measuring magnetic quantities				•	210
5.10.	Means for measuring magnetic quantities					214
5.1/.	raphy					221
Bibliog	rapny	•	•		-	
	a de manda esta de Carte de la Re					
Section	6. General Aspects of the Utilization of Electrical Eq	laı	pment			
6.1.	Standardization in the electrical engineering industry.	,	UTILLE	atı	.on	
	of the System of State Standards	•		•	•	222
	Categories and types of standards			•	•	223
	Standardization of electrotechnical products					224
	Quality certification of electrotechnical products .	_				225
	Utilization of ranges of preferable numbers in electric	1و،	engir		·_ ·	
	Utilization or ranges of preferable numbers in electric	- CL.L	g11	1		228
	ing	•				
6.2.	Major parameters of electric networks and the equipment	c	onneci	:ea	τo	
	them	•		٠.	•	231
	Concrete					

	Ranges of nominal direct and alternating voltages	231
	Nominal frequencies	232
	Range of nominal currents	233
	Ranges of basic parameters for electrical machinery	233
	Nominal outputs of power transformers	234
	Ranges of nominal linear dimensions	235
6.3.	General requirements for safety and for protecting electrical	
	equipment from external influences	236
	Electrical equipment safety	236
	Degrees of protection of electrical equipment provided by sheathing.	236
	Degrees of protection of electrical machinery	237
	Degrees of protection of electrical apparatus	237
	General aspects of the effects of mechanical and climatic environ-	231
	mental factors upon electrical equipment under operational con-	220
	ditions	238
	Specifications regarding climatic effects	239
	Specifications regarding mechanical effects	243
	Electrical equipment for regions with tropical or cold-weather	
	climates	244
	Testing electrical equipment for resistance to the effects of cli-	
	matic and mechanical factors of the environment	245
6.4.		246
6.5.		247
	System of standardization of terms and definitions in electrical	247
	engineering	247
	Letter designations in electrical engineering	240
0	7 Assessment to the DCC of the Second of New Market 1 and	
	7. Ascertaining the Effectiveness of New Technology	
7.1.		249
	electrical products	249
	Place and role of calculations of economic effectiveness during the	010
	development of new electrical products	249
	Determining the annual economic impact	250
	Selecting bases for comparison and conditions for contrasting	
	variants	250
	Calculation of the time factor	251
7.2.	Determining expenditures for the manufacture and operation of elec-	
	trical equipment	251
	Composition of capital investment and current expenditures for the	
	manufacture of new electrical equipment	252
	Cost of manufacturing electrical equipment	252
	Determining expenditures for the operation of electrical equipment .	256
7.3.	Limiting costs of and allowances for the quality of electrical pro-	
	ducts	257
	Calculation of the limiting costs for a new electrical product	257
	Calculation of the cost allowances for a new product	257
	Stepped costs. Maximum price	258
7.4.		259
	Technical-economic indicators of compared versions of electrical	
	a modulate	250

## 57

	Example for determining the annual economic impact due to the intro- duction into the economy of a new electric motor	259
	Determining the effectiveness of new electrical products on the	
	basis of the graduated-index method of evaluation	260
	Procedure for economic optimization of electrical products being	
7.5.	Procedure for economic optimization of electrical products seems	261
	designed	261
	Economic optimization matrix.	201
	Methods for determining the significance of the indicators and	261
	measures for improving the effectiveness of electrical products .	
	Functional-cost analysis	262
7.6.	Features of determining the economic impact of measures taken to	06.0
	improve production methods	262
Bibliog:	improve production methods	263
	a c	
	8. Safety of Electrical Installations and Equipment	263
8.1.	General information on electrical safety	263
	Basic concepts and definitions	264
	Effect of electric current on the human organism	264
	Classification of electrical installations	265
8.2.	Electrical equipment safety (safety engineering)	268
8.3.	Protective measures in electrical installations	268
	Grounding and neutral-wire grounding in electrical installations.	
	Grounding in electrical installations of over 1,000 V · · · ·	269
	Grounding in electrical installations of up to 1,000 V	270
	Calculating the ground	272
	Grounding connections	276
	Grounding and protective neutral conductors	276
	Grounding and neutral-wire grounding of portable receptacles	277
	Protective disconnect	277
8.4.	Protection from electrostatic discharge	279
0.4.	courses of electrostatic discharge during production	279
	Fliminating discharge hazards	280
8.5.	Floatwicel equipment in fire-hazard areas and Installations	282
0.5.	Classification of types of production and areas with respect to fire	
	harrard	282
	Selection of equipment for fire-hazard areas	283
	Classification of materials and designs with respect to the degree	
	of combustibility	283
	Fire precaution and the extinguishing of fires.	286
	graphy	287
B161108	graphy	
Section	9. Electrical-Insulation Materials	
9.1.	Basic definitions and classification of electrical-insulation	
,	materials	287
9.2.	Heat resistance of electrical insulating materials	288
9.3.	Liquid and semiliquid insulators	292
9.4.	Organic polymer insulators	293
9.5.	Flectrical-insulating papers and cardboards	297
9.6.		300
	Electrical-insulating compounds.	304

58

## FOR OFFICIAL USE ONLY

9.8.	Electrical-insulating varnished fabrics	•	• 1		•	308
9.9.		•	• •			309
9.10.	Mica-based electrical-insulating materials					310
9.11.	Electroceramic materials		. ,			315
Bibliog:	Electroceramic materials	•	• (		•	315
Section	10. Semiconducting Materials					
10.1.	General information					316
10.2.	Obtaining semiconducting materials	•	•	• •	•	316
	Basic parameters of semiconducting materials and metho		for	· ·		310
10.5.	application	us	TOL	tile.	TT	316
10.4	Electrophysical parameters of semiconductors	•	•	• •	•	
Riblian	confirmation parameters of semiconductors	•	•	• •	•	319
DIDITOR	caphy	•	•	• •	•	325
Section	11. Magnetic Materials					
11.1.	The second secon	out	the	eir		
	parameters and characteristics					325
11.2.	Classification of magnetic materials					328
11.3.	Magnetic materials for operations within a broad range	of	apr	olica	1-	
	tions of magnetic induction at low and elevated freq	uen	cie	3 .		333
11.4.	Magnetic materials for operations in weak magnetic fie	lds			•	337
11.5.	Magnetic materials with a square hysteresis loop				•	341
11.6.	Special-purpose magnetically soft materials				•	344
11.7.	Magnetic materials for permanent magnets	•	•	•	•	346
11.8.	Costs	•	•	•	•	
	caphy	•	•		•	350
		•	•	•	•	330
Section	12. Conducting Materials, Noninsulated Conductors and	Bus	es			
12.1.					_	351
12.2.	Copper	-			•	352
12.3.	Brass				•	355
12.4.	Conducting bronzes	-			•	359
12.5.	Aluminum	•			·	363
12.6.	Aluminum alloys	•		•	•	365
12.7.	High-resistance alloys for electrical measuring device			•	•	367
12.8.	High-resistance heat-resistant alloys		•	•	•	371
12.9.	Superconducting and cryoconducting materials	•		•	•	375
12.10.	Contact materials and electrical brushes	•	• •	•	•	377
12.11.	Copper wiring	•	• •	•	•	384
12.12.	Aluminum and aluminum-alloy wiring	•	• •	•	•	385
12.13.	Noninsulated wiring	•	• •	•	•	387
12.14.	Buses and bonding strips	•	• •	•	•	391
12.15.	Contact conductors	•		•	•	391
12.16.	Sections for commutators of electrical machinery.	•	• •	•	•	394
	Costs	•		•	•	205
Ribliogr	Costs	•	• •	•	•	395
PIDITORI	арпу	•	• •	•	•	396
Section	13. Electrical-Insulation Designs and Insulators					
13.1.	General requirements for electrical-insulation designs				•	397
	General information					397
	Internal and external inculation					200

	Requirements for dielectric strength of insulators .					•	399
	Dielectric strength of standard electrical-insulation	gap	8				401
13.2.	General information	9F	_				401
	General information		•				401
	Gas gaps	-	_				403
	Insulating gaps in oil						404
	Gaps in oil along the surface of a solid dielectric.			•			406
	Caps in oil along the surface of a solid dielectric.	•	-				406
	Paper-oil insulators	•	•				407
			_	_			40/
13.3.	Insulator calculations	•	•	•	•		407
	Insulator calculations	•	•	•	•		407
	Calculation of the mechanical strength of insulators	•	•	•	•	•	409
	Calculation of the mechanical strength of Laborator Calculation of a base insulator	•	•	•	•	•	410
	Calculation of a bushing insulator	•	•	•	•	•	411
13.4.							
	n to the standard of the stand	•	•	•	•	•	
	70 Januari a bases	•	•	•	•	•	
		•	•	•	•	•	
13.5.	Lead-ins for 100-kV lines and higher.	•	•	•	•	•	420
13.6.		•		•	•	•	
		•.	•	•	•	•	7 20
ibliog	raphy	•	. •	• .	•	•	422
	• •						
Section	14. Cable Equipment						. 422
14.1.	14. Cable Equipment Classification of cable equipment	•	•	•	•	•	422
14.2.							
14.3.			-		•	•	730
14.4.	Weakeds of lowing nower cable	•	•	•	•	•	437
14.5.							445
14.6.	n	•	•	•	•	•	
14.7.		•	. •	•	•	•	4 54
14.8.	Power and installation wiring and connecting cords .	•	•	•	•	•	464
14.9.	Unal un ablac and Wires	•	•	•	•	•	404
14.10.							
14.11.	Warney with enamel-fiber, paper and film insula	CTON	ι.	•	•	•	472
	vizina malah faharalaga ingilation	•	•	•	•	•	474
14 13	Other tunes of magnet Wire	•	•	•	•	•	475
	o	•	•	•	•	•	477
14.14. 1461.100	graphy	•	•	•	•	•	480
Cootion	15. Semiconductor Instruments and Microcircuitry						
15.1.	a to the dada	•		•	•	•	481
15.1.			•		•	•	483
1).2.	Transistors		•	•	٠	•	483
	Field-affect transistors.		•		٠.	•	485
	Fauluations directing of himolar translators	•	•	•	•	•	
15.3	Thyristors						490
1 1 1	INVERSIOUS						

## FOR OFFICIAL USE ONLY

15.4.	Electroopt	tical	devi	lces		•	•		•			•	•	•			•			494
	Barrage pl	notoce	e11s	and	pho	tod:	iode	es												494
	Phototrans	sistor	cs .		•		•													494
	Opto-isola	ators																		495
15.5.	Integrated	d circ	uite	з.														•		495
15.6.	Operations	al amp	lifi	lers	and	th	eir	apı	olio	cat	ion			•	-		•	•	•	497
	Basic para	meter	s of	ODe	erat	iona	n 1 :	ສຫກ	1 1 f	ier	R.	•		·	·	•	•	•	•	497
	Feedback	in ope	rati	iona	Lam	n1ii	fie	rs			•	•	:	•	•	•	·	•	•	498
	Summation	of vo	ltas	es.					Ī	•	•	•		•	•	•	•	•	•	499
	Integration	on of	volt	300	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	499
	Stabilizat	tion of	of vo	il ta	780	•	•	•	•	•	•	•	•	•	•	•	•	•	•	499
	Active fil	tere	,	, , , , ,	500	•	•	•	•	•	•	•	•	•	•	•	•	•	• .	499
	Voltage li	initer	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	
	Comparator		• •											•	•	•	•	•	•	500
	Square-way											•	•	•	•	•	•	•	•	500
	Voltace to	ve voi	Lage	ger	iera	COL	s.		•	•		٠,	•	•	•	•	•	•	•	500
15.7.	Voltage-to	)-1 rec	luenc	усс	nve	rte	r8 a	ana	au	rat:	ıon	.oi	pu	Tse	•	•	•	•	•	501
13.7.	Integrated	1 Tog	LC C 3	.rcu	LES	and	.tn	eır	apı	) T 10	cat:	Lon	•	•	•	•	•	•	•	501
	Basic cond	epts	oi 1	.0 <b>g1</b> 0	: tu	nct:	ion	s.	•	• .	•	•	•	•	•	•		•	•	501
	Relations	or To	gic	alge	bra	•	. •	•	•	•	•	•	•	•	•	•	•	•	•	502
	Minimizati	ion of	log	gic e	xpr	ess:	ion	8.	•	•	•	•	•	•.	•	•	•			503
	Examples of	of the	e des	cri	tio	n ai	nd 1	nin	imi	zat:	ion	οf	10	gic	ex	pre	ssi	ons		503
	Hardware i		ien ta	tion	of	10	gic	re.	lat:	ions	в.	•		•	•		•		•	505
	Decoders .			•										•	•	•			•	506
	Flip-flops	and	flip	-f1	op d	evi	ces								•				.•	506
	Counters .																		•	508
	Registers.					•			•											508
	Distributo	rs.												•					•	509
	Synthesis	of me	mory	ci	cui	ts													_	509
	Integrated	circ	uits	wit	:h a	hi	zh d	deg	ree	of	int	ee:	rat	ion	an	d m	icr	o~	•	
	processo																	•	_	510
Bibliogra						-		-	_	-	:	•	•		•	•	•	•	•	511
		•	-	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•.	711

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## ELECTRIC POWER-SUPPLY SOURCES WITH SPECIAL CHARACTERISTICS

Kiev ISTOCHNIKI ELEKTROPITANIYA SO SPETSIAL'NYMI KHARAKTERISTIKAMI in Russian 1979 (signed to press 6 Feb 79) pp 2, 162-3, 165-71

[Annotation, table of contents and list of abstracts from book "Electric Power-Supply Sources With Special Characteristics", edited by N. I. Sukhomlinskaya, Izdatel'stvo "Naukova dumka", 1,500 copies, 172 pages]

[Text] This collection includes papers devoted to the development and investigation of electromagnetic and semiconductor devices for the regulation and stabilization of current, voltage, power and other parameters influencing the working capacity and efficiency of electrical and electronic equipment. This collection covers questions of the theory and practical application of regulators of electrical quantities as applied to loads having variable parameters.

This collection is intended for scientific and engineering workers specializing in the fields of transformer engineering and automation equipment.

Contents	Page
Volkov, I. V., Gubarevich, V. N., Isakov, V. N., and Kaban, V. P. Inductive-capacitive converters with regulated amplitude and constant phase for the output current	3
Volkov, I. V., and Smolyanskiy, I. I.  The application of wave matrices in the study of inductive-capacitive converters	11
Gubarevich, V. N., Isakov, V. N., and Kaban, V. P. Two-element circuits for inductive-capacitive converters	. 17
Gubarevich, V. N., Isakov, V. N., and Kaban, V. P.  II-type circuits for three-phase inductive-capacitive converters	28
Kalinin, V. N., and Kusnetsov, O. G. Calculation of the spectral composition of nonsinusoidal currents in an inductive-capacitive converter using a digital computer	39

## 62

Kusnetsov, O. G., and Kalinin, V. N.  Power relations among components of polyharmonic oscillations arising in a single-phase inductive-capacitive converter	46
Lipkovskiy, K. A., and Aleksandrov, M. M. Structures of composite power-supply systems	51
Zakrevskiy, S. I., and Basan'ko, Yu. V.  Effect of parameter variations on the operation of an inductive- capacitive converter in the constant current-consumption mode	57
Samcheleyev, Yu. P., Obukhov, S. G., Kalyuzhnyy, V. V., and Komarskiy, V. V.	
Controlled-velocity direct-current electric drive supplied from a parametric power source	62
Samcheleyev, Yu. P., Zhilyakov, V. I., Kalyuzhnyy, V. V., and Chernoivan, V. P.	
Synthesis of automatic velocity-control systems when a direct current motor is supplied from a power source	66
Zozulev, V. I., Gorbachev, M. N., Zakrevskiy, S. N., and Inyakin, I. A. Analysis of the electromagnetic processes in a three-phase series-type regulator	72
Dovgalevskiy, A. Yu., Starodumov, Yu. I., and Tsymbal, K. K. Two-terminal current regulators in circuits with high-voltage power leads	83
Grechko, E. N., Adamishin, M. M., Boyko, A. P., and Kutovoy, V. I.  Control of a three-phase asynchronous motor based on impulse duration modulation of alternating voltage curves	89
T 17 3 T	
Pentegov, I. V., and Legostayev, V. A.  Comparison of the technical-economic characteristics of and limitations on the application of reservoir reactors with and without steel	97
Sarv, V. V.  Improving the quality of the output current from gate-type converters through the use of a regulated inductive power-storage circuit	106
Spirin, V. M.  Simplified equation for designing reactors with a linear voltage- amperage characteristic	113
Tonkal', V. Ye., Bezgachin, N. I., and Nikitin, V. B.  Power characteristics of reactive two-terminal networks	. 118
Kuznetsov, V. G., and Ugrimov, S. F.	
Inductive-capacitive circuit for zero-series current compensation with nonsymmetry in the power-supply voltage	130

## 63

Mostovyak, I. V. Compensation of a three-phase system with a nonsymmetrical power source	135
Starodumov, Yu. I. Principles for constructing devices for igniting gas discharges	139
Andriyevskiy, Ye. A., and Sheverdin, G. P.  Pulse magnetizing device with stabilization of the discharge current  amplitude and the time for charging the capacitor bank	146
Spirin, V. M., Krynko, N. V., and Kurach, A. M.  Effect of power-supply voltage fluctuations on the pulse recurrence rate of pulse-current generators	151
Pavlov, V. B., Kravchenko, Yu. I., and Yakimov, O. S. Study of thyristor converters with parallel capacitive switching	158

UDC 621.372.061

INDUCTIVE-CAPACITIVE CONVERTERS WITH REGULATED AMPLITUDE AND CONSTANT PHASE FOR THE OUTPUT CURRENT

[Abstract of article by Volkov, I. V., Gubarevich, V. N., Isakov, V. N., and Kaban, V. P.]

[Text] This article presents circuits of three-phase inductive-capacitive converters whose regulating element is a choke with a rotating magnetic field. It is shown that the circuits described make it possible to employ a smooth contactless method of regulating the level of stabilized current in a load while keeping its phase unchanged. Areas for the practical application of the circuits studied are determined. 6 illustrations, 6 titles in bibliography.

UDC 621.372.061

THE APPLICATION OF WAVE MATRICES IN THE STUDY OF INDUCTIVE-CAPACITIVE CONVERTERS

[Abstract of article by Volkov, I. V., and Smolyanskiy, I. I.]

[Text] The article examines questions of the application of wave-matrix apparatus in the study of multiphase inductive-capacitive converters. It assesses the capabilities of inductive-capacitive converters in connection with their power characteristics and the application of the given apparatus. It also cites conditions under which these characteristics are optimal.

Further approaches are outlined for the effective utilization of wave matrices in the study of inductive-capacitive converters. 1 illustration, 8 titles in bibliography.

UDC 621.372.061

TWO-ELEMENT CIRCUITS FOR INDUCTIVE-CAPACITIVE CONVERTERS

[Abstract of article by Gubarevich, V. N., Isakov, V. N., and Kaban, V. P.]

[Text] The article provides a comparative analysis of two-element circuits for inductive-capacitive converters. General mathematical relationships are obtained for converters of L-type structure which make it possible to carry out the best possible selection of this or that circuit when designing specific equipment with given specificiations. 3 illustrations, 4 tables, 1 title in bibliography.

UDC 621.372.061

II-TYPE CIRCUITS FOR THREE-PHASE INDUCTIVE-CAPACITIVE CONVERTERS

[Abstract of article by Gubarevich, V. N., Isakov, V. N., and Kaban, V. P.]

[Text] The article analyzes various methods of connecting the phase e.m.f. of a three-phase power supply network to inductive-capacitive converters of II-type structure. Formulas are obtained for calculating the parameters of the current source. Recommendations are given regarding the selection of the most efficient circuits. 5 illustrations, 3 tables, 3 titles in bibliography.

UDC 621.372.061

CALCULATION OF THE SPECTRAL COMPOSITION OF NONSINUSOIDAL CURRENTS IN AN INDUCTIVE—CAPACITIVE CONVERTER USING A DIGITAL COMPUTER

[Abstract of article by Kalinin, V. N., and Kusnetzov, O. G.]

[Text] This article examines the use of a digital computer to calculate the amplitude, frequency and phases of the harmonic components of a nonsinusoidal function with regard to its discrete spectrum over a finite interval of time. An algorithm is suggested and its flow-chart is presented. Results obtained are illustrated by examples of calculation. 2 illustrations, 2 titles in bibliography.

UDC 621.372.061

POWER RELATIONS AMONG COMPONENTS OF POLYHARMONIC OSCILLATIONS ARISING IN A SINGLE-PHASE INDUCTIVE-CAPACITIVE CONVERTER

[Abstract of article by Kusnetzov, O. G., and Kalinin, V. N.]

[Text] Based on power relationships, this article proves the existence of steady-state nonsinusoidal oscillations in whose spectrum at amplitude predominate the harmonics of frequencies  $\omega$  and  $\omega + \Delta \omega$ , where  $\omega$  is the frequency of the converter's sources of e.m.f. 1 illustration, 2 titles in bibliography.

65

UDC 621.314.572

STRUCTURES OF COMPOSITE POWER-SUPPLY SYSTEMS

[Abstract of article by Lipkovskiy, K. A., and Aleksandrov, M. M.]

[Text] The article examines the construction of power-supply systems which are various combinations of voltage and current sources. Structures are analyzed for composite power-supply systems and conditions are ascertained in each of the combinations which make it possible to stabilize the load current. 2 illustrations, 1 table, 4 titles in bibliography.

UDC 621.372.061

EFFECT OF PARAMETER VARIATIONS ON THE OPERATION OF AN INDUCTIVE-CAPACITIVE CONVERTER IN THE CONSTANT CURRENT-CONSUMPTION MODE

[Abstract of article by Zakrevskiy, S. I., and Basan'ko, Yu. V.]

[Text] This article analyzes the operation of a three-phase bridge-type circuit for an inductive-capacitive converter in the constant consumption-current mode with variation in the circuit element parameters and the supply-voltage frequency. 4 illustrations, 4 titles in bibliography.

UDC 62.83.001.5

CONTROLLED-VELOCITY DIRECT-CURRENT ELECTRIC DRIVE SUPPLIED FROM A PARAMETRIC POWER SOURCE

[Abstract of article by Samcheleyev, Yu. P., Obukhov, S. G., Kalyuzhnyy, V. V., and Komarskiy, V. V.]

[Text] Principles are examined for the construction of reversible and nonreversible electric drives when the DC motor's armature is supplied from a parametric current source.

The suggested drive systems possess high dynamic and static indicators, simplicity of circuit design and reliability. Control is accomplished at the motor's armature circuit. 2 illustrations, 4 titles in bibliography.

UDC 62.83.001.5

SYNTHESIS OF AUTOMATIC VELOCITY-CONTROL SYSTEMS WHEN A DIRECT CURRENT MOTOR IS SUPPLIED FROM A POWER SOURCE

[Abstract of article by Samcheleyev, Yu. P., Zhilyakov, V. I., Kalyuzhnyy, V. V., and Chernoivan, V. P.]

[Text] This article synthesizes the structure of a DC electric drive when the current is provided by a parametric current source. Results are presented for a

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study of a drive based on a mathematical model and an actual unit. The electric drive possesses high response speed and simplicity of circuit design. 5 illustrations, 5 titles in bibliography.

UDC 621.314.722

ANALYSIS OF THE ELECTROMAGNETIC PROCESSES IN A THREE-PHASE SERIES-TYPE REGULATOR

[Abstract of article by Zozulev, V. I., Gorbachev, M. N., Zakrevskiy, S. N., and Inyakin, I. A.]

[Text] An analysis is given of the electromagnetic processes for periodic operation of a three-phase preresonant series-type voltage (current) regulator. Expressions are derived for the instantaneous values and the voltage and current waveforms at the regulator output are ascertained. 5 illustrations, 4 titles in bibliography.

UDC 621.316.721.1:621.382

TWO-TERMINAL CURRENT REGULATORS IN CIRCUITS WITH HIGH-VOLTAGE POWER LEADS

[Abstract of article by Dovgalevskiy, A. Yu., Starodumov, Yu. I., and Tsymbal, K. K.]

[Text] This article examines circuit designs for two-terminal current stabilizers possessing high values for the output resistance. Mathematical expressions are found and the characteristics of the stabilizers are analyzed. Experimental data are cited. 6 illustrations, 4 titles in bibliography.

UDC 621.314.26

CONTROL OF A THREE-PHASE ASYNCHRONOUS MOTOR BASED ON IMPULSE DURATION MODULATION OF ALTERNATING VOLTAGE CURVES

[Abstract of article by Grechko, E. N., Adamshin, M. M., Boyko, A. P., and Kutovoy, V. I.]

[Text] This article conducts a harmonic analysis of the alternating voltage curves processed through impulse duration modulation. The basic mathematical relations are derived. A thyristor device is described which controls an asynchronous electric drive and which realizes the given form of impulse duration modulation in the starting and stopping modes of operation. 2 illustrations, 3 tables, 3 titles in bibliography.

UDC 621.3:621.791.75

COMPARISON OF THE TECHNICAL-ECONOMIC CHARACTERISTICS OF AND LIMITATIONS ON THE APPLICATION OF RESERVOIR REACTORS WITH AND WITHOUT STEEL

[Abstract of article by Pentegov, I. V., and Legostayev, V. A.]

[Text] Questions are examined regarding the optimal geometry for reactors. Formulas are developed for determining their optimal geometric parameters. The article

compares reactors with and without steel on the basis of weight and the cost of the core materials. Power engineering areas for the most profitable application of these types of reactors are indicated. 4 illustrations, 9 titles in bibliography.

UDC 621.314.061

IMPROVING THE QUALITY OF THE OUTPUT CURRENT FROM GATE-TYPE CONVERTERS THROUGH THE USE OF A REGULATED INDUCTIVE POWER-STORAGE CIRCUIT

[Abstract of article by Sarv, V. V.]

[Text] It is shown that the operating efficiency of an inductive power-storage circuit can be increased substantially and the spectral composition of the input and output currents of gate-type converter circuits can be improved while rhythmically varying the inductance of the storage circuit. This article examines the suitability of the proposed method in an alternating voltage regulator, in a rectifier and in a current inverter. 5 illustrations, 10 titles in bibliography.

IDC 621.318.434.1

SIMPLIFIED EQUATION FOR DESIGNING REACTORS WITH A LINEAR VOLTAGE-AMPERAGE CHARACTERISTIC

[Abstract of article by Spirin, V. M.]

[Text] A method is described for deriving a simplified equation for designing reactors with a linear voltage-amperage characteristic. The method is based upon the realization of the full factorial experiment 2<sup>3</sup> with the conversion of independent variables. An example is provided for deriving a simplified equation for the design of reactors for a specific case.

UDC 621.372.061

POWER CHARACTERISTICS OF REACTIVE TWO-TERMINAL NETWORKS

[Abstract of article by Tonkal', V. Ye., Bezgachin, N. I., and Nikitin, V. B.]

[Text] Questions are examined regarding the application of characteristic impedence and characteristic admittance functions for the analysis and synthesis of alternating current power circuits containing reactive elements. Four classes of two-terminal networks are analyzed with respect to the total output of the reactive elements. Features relating to the application of reactive and characteristic impedence and admittance are examined when calculations are performed on a digital computer. 3 illustrations, 1 table, 4 titles in bibliography.

UDC 621.317.619

INDUCTIVE-CAPACITIVE CIRCUIT FOR ZERO-SEQUENCE CURRENT COMPENSATION WITH NONSYMMETRY IN THE POWER-SUPPLY VOLTAGE

[Abstract of article by Kusnetzov, V. G., and Ugrimov, S. F.]

68

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[Text] An analysis is made of the influence of voltage nonsymmetry and unbalance as well as of the variation in the element parameters in the zero-sequence inductive-capacitive current-compensation circuit due to calculated values. Formulas are presented which make it possible to assess the influence of these factors upon the current in positive and negative sequences during total and partial compensation of the zero-sequence current. 3 illustrations, 4 titles in bibliography.

UDC 621.326

COMPENSATION OF A THREE-PHASE SYSTEM WITH A NONSYMMETRICAL POWER SOURCE

[Abstract of article by Mostovyak, I. V.]

[Text] This article establishes conditions for balancing a three-phase system with a nonsymmetrical load. The nature of the balancing is determined. 2 titles in bibliography.

UDC 621.377.534.032.43

PRINCIPLES FOR CONSTRUCTING DEVICES FOR IGNITING GAS DISCHARGES

[Abstract of article by Starodumov, Yu. I.]

[Text] This article axamines various devices for igniting a gas discharge. The advantages and disadvantages of each are analyzed. Their areas of application are determined. Ignition systems are classified. 1 illustration, 9 titles in bibliography.

UDC 621.311.6:621.317.44

PULSE MAGNETIZING DEVICE WITH STABILIZATION OF THE DISCHARGE CURRENT AMPLITUDE AND THE TIME FOR CHARGING THE CAPACITOR BANK

[Abstract of article by Andriyevskiy, Ye. A., and Sheverdin, G. P.]

[Text] This article describes a pulse magnetizing device with stepped variable voltage supply to the capacitor bank and stabilization of the discharge current amplitude with variation in the parameters of the charging and discharging circuits and in the charge time of the pulse capacitor bank.

Mathematical expressions are presented which make it possible to select parameters for the circuit elements which will insure the stability of the discharge current and the charge time of the capacitor bank. I illustration, 4 titles in bibliography.

UDC 621.373.072.6

EFFECT OF POWER-SUPPLY VOLTAGE FLUCTUATIONS ON THE PULSE RECURRENCE RATE OF PULSE-CURRENT GENERATORS

[Abstract of article by Spirin, V. M., Krynko, N. V., and Kurach, A. M.]

69

[Text] This article provides an analysis of the influence of voltage oscillations in the power supply upon the pulse recurrence rate in pulse current generators of electrohydraulic installations. For pulse current generator circuit designs with inductive reactance in the primary alternating current circuit and with inductive-capacitive conversation of the voltage source to a current source, variation in the pulse recurrence rate is specified when there are fluctuations in the supply network voltage. 3 illustrations, 7 titles in bibliography.

UDC 621.382.2.014.2

STUDY OF THYRISTOR CONVERTERS WITH PARALLEL CAPACITIVE SWITCHING

[Abstract of article by Pavlov, V. B., Kravchenko, Yu. I., and Yakimov, O. S.]

[Text] Variants of circuit designs for thyristor direct-current interrupters. Their operational characteristics at low input voltage are shown. The circuit designs presented are calculated on a computer and the results of the machine computation are compared with the experimental results. A deduction is made regarding the unsuitability of shunting a power thyristor with a circuit of inductance and a diode when supplying the interrupter with parallel capacitive switching from a low-voltage source. 2 illustrations, 1 table, 3 titles in bibliography.

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9512

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UDC 621.371

#### EXPERIMENTAL STUDIES ON RADIO WAVE PROPAGATION

Moscow EKSPERIMENTAL'NOYE ISSLEDOVANIYE RASPROSTRANENTYA RADIOVOLN in Russian 1980 pp 2, 150-151

[Annotation and table of contents from book "Experimental Studies on Radio Wave Propagation", by V. Ye. Kashprovskiy, Institute of Terrestrial Magnetism, the Ionosphere and Radio Wave Propagation of the USSR Academy of Sciences, Izdatel'stvo'Nauka"]

[Text] On the basis of the analysis of propagation processes and properties of receiving antennas, the author studies the effects of their characteristics on the registered values of field intensity and the degree of their applicability in experiments. Modern methods of measuring low voltages are examined, fundamentals of the computation of dummy antennas are given, modern field recording instruments are described, and ways of increasing the accuracy of measurement are mapped out.

It is shown that multibeam propagation mechanisms impair the quality of radio channels and limit the volumes of information about propagation processes. The book gives an idea of the methods of oblique sounding and organization of measurements of the intensity of individual beams, their number, and angles of arrival at the reception point. Methods for measuring three components of the wave field are outlined. Methods for determining parameters of soils information on which is necessary for calculating antenna characteristics are presented. Recommendations are given for the planning of systems of oblique sounding of the ionosphere and systematic observations of propagation processes.

This monograph is intended for scientists and engineers, and for graduate and undergraduate students of vuzes engaged in studying propagation processes of radio waves controlling the work of real radio channels, as well as for designers of such instruments.

Figures -- 89, tables -- 2, bibliography -- 91 items.

Contents	page
Introduction	3
Chapter I. Propagation of Long, Medium, and Short Radio Waves. Structure of Wave Fields 1. Some General Considerations 2. Propagation of "Earth" Waves	5 5 5

71

3.	Propagation of Waves of Medium and Long Wave Ranges Through the Ionosphere	12
4.	Mechanisms of the Propagation of Radio Waves of the Short-Wave Range	15
1. 2. 3.	r II. Receiving Antennas as Elements of Field Intensity Measurers Frame Antenna and Its Basic Properties Vertical Asymmetrical Dipole (Rod Antenna) Horizontal Dipole and Its Basic Properties as a Measuring Antenna Characteristics of a Traveling-Wave Antenna	19 19 27 34 40 46
5.	Log-Periodic Antennas and Their Peculiarities	40
1. 2. 3.	r III. Organizational Problems of Intensity Measurements of Wave Fields Classification of Field Intensity Measurements Minimal Dimensions of the Wave Zone Area of Earth Surface Actively Participating in the Formation of Wave Fields Influence of Local Objects Antenna Systems of Receiving Centers and Their Interaction	49 49 51 53 56 67
Chapte	r IV. Fundamentals of Measuring the Electromotive Force of Receiving Antennas and Field Intensities Corresponding to Them	70
1.	Differences in the Conditions of Field Measurements	70 71
2.	Basic Principles of Wave Field Measurement	74
3.	Substitution Method Calibration Method	76
	Method of Antenna Short-Circuiting	78
5. 6.	Equivalents of Antenna-Feeder Systems and Their Realization	80
7.	Problems of Continuous Recording of Field Intensity	85
8.		91
Chapte	er V. Field Intensity Measurement as a Means for Determining the Parameters of a Radio Channel	95
1.	Long-Wave Ranges	96
2.	Long-Wave Ranges	101
3.	Shortwave Range and the Control of the Parameters of Dummy Antennas	104
4.	Field Intensity Measurement for Evaluating the Characteristics of the Directivity of Receiving and Transmitting Antennas	107
5.	Measurement of Soil Parameters Along the Propagation Route of	112
_	Earth Waves (Route Conductivity of Soils OT)	114
6. 7.	a - t wee Daws-bone of Coile in the Shorte	115
Chapt	er VI. Ways of Improving the Methods and Techniques of Measuring Wave Fields	120

72

10,233 CSO: 1860/221

## DNLY

<ol> <li>Specific Requirements in Measuring Wave Fields</li> </ol>	120
2. Determination of the Main Properties of Propagation Mechanisms	124
3. Possible Principles of the Organization of an Observation	
System of Ionospheric Situation and Effective Control of Radio Communication System	141
Conclusion	144
Bibliography	147
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73

UDC 620.17

FUNDAMENTALS OF THE ELECTROMAGNETIC METHOD OF STRESS ANALYSIS IN ANISOTROPIC MEDIA

Kiev OSNOVY ELEKTROMAGNITNOGO METODA ISSLEDOVANIYA NAPRYAZHENIY V ANIZO-TROPNYKH SREDAKH in Russian 1980 (signed to press 3 Oct 79) pp 2-4

[Annotation and table of contents from the book "Fundamentals of the Electromagnetic Method of Stress Analysis in Anisotropic Media", by Ivan Petrovich Vasil'chenko and Bogdan Lyubomirovich Pelekh, Izdatel'stvo "Naukova dumka", 1150 copies, 116 pages]

[Text] This monograph presents a method for determining stress-strain conditions in anisotropic media through the use of the non-visible region of the electromagnetic spectrum. An exposition of the theoretical basis of the electromagnetic method for determining stress in anisotropic media is provided. Possibilities for applying this method to dynamic problems for anisotropic bodies are indicated. State-of-the-art implementation of the method is described. Functional diagrams of experimental devices are also provided, as well as results of the investigations.

The work is intended for the use of scientific and technical engineering personnel engaged in experimental investigations of the strength of solid deformable bodies.

#### Table of Contents

Foreword		,
Introducti	on	9
	Analysis of stress-strain conditions in anisotropic model materials using the visible-light region of the electromagnetic spectrum. Current status of the problem Extension of photomechanical methods to non-transparent materials using the microwave region of electromagnetic waves	9

74

Chapter 1.	Theoretical Bases of the Electromagnetic Method for Determining Stress in Anisotropic Media	16
2. 3.	Physical bases of the method Some data from the theory of electromagnetic wave propaga- tion in electrically anisotropic dielectric media Relationship of the phase difference of polarized waves to the strain state Passage of a polarized electromagnetic wave through a stressed plate. Intersecting and parallel planes of polarization	16 18 21
5.	Basic correspondences which determine the relationship between the strain state of a medium and the electromagnetic characterisitics of a polarized wave	21
1. 2. 3. 4.	Apparatus and Technical Applications of the Electromagnetic Method Radiopolarization equipment Radiointerferometry Characteristics of stress analysis in planar plastic models using millimeter waves of microwave region Characteristics of measurements made in the visible-light and infrared ranges on full-scale fiberglass products Stress distribution analysis using the millimeter region of electromagnetic waves Concerning the determination of dynamic stresses	31 32 35 . 38 . 42 . 45
1.	Acoustic Methods for Investigating the Physical-Mechanical Properties of Anisotropic Reinforced Plastic Introductory remarks Resonance methods for determining the elastic properties of composite materials Some problems in the measurement of mechanical properties at high temperatures	53 53 51 63
1. 2. 3. 4.	Analysis of the stress-strain condition of anisotropic media in the visible-light range(method dealing with the photoelasticity of anisotropic bodies)  Obtaining piezo-optical material with anisotropic elastic properties for use in optical modeling  Strain concentration in orthotropic perforated plates  Analysis of granular structures  "Freezing" of strains in the analysis of thick lamellate reinforced plastic by the polarization-optical method  Assessing the effect of Poisson's ratio on the stress state in the solution of volumetric problems by the freezing method  Stress state analysis in thick-walled conical shells	70 73 76 80 89
Bibliograph	hy	109
COPYRIGHT:	Izdatel'stvo "Naukova dumka", 1980	
9481 CSO: 1860/	197	

75

UDC 621.396.932(031)

HANDBOOK OF SHIPBOARD RADIO COMMUNICATION AND RADIO NAVIGATION EQUIPMENT. VOL. 2: RADIO NAVIGATION EQUIPMENT

Leningrad SPRAVOCHNIK PO SUDOVOMU OBORUDOVANIYU RADIOSVYAZI I RADIONAVIGATSII.

TOM 2: OBORUDOVANIYE RADIONAVIGATSII in Russian 1979 (signed to press 13 Nov 79)

pp 2-4, 230-31

[Annotation, foreword and table of contents from book "Handbook of Shipboard Radio Communication and Radio Navigation Equipment", by Aleksandr Mustafovich Bayrashevskiy, Yuriy Yeliseyevich Gornostayev, Aleksandr Vasil'yevich Zherlakov, Aleksandr Anatol yevich Il'in, Oleg Vasil'yevich Kononov and Nikolay Timofeyevich Nichiporenko, Izdatel'stvo "Sudostroyeniye", 12,000 copies, 232 pages]

#### [Text]

#### Annotation

The second volume of this handbook contains basic information on modern radio navigation and radar equipment for seagoing vessels. A classification of the devices is given along with their characteristics, and the makeup and composition of the equipment are given along with functional diagrams. The basic requirements for installing and operating the equipment under shipboard conditions are explained.

This handbook is intended for a wide group of engineering and technical workers in design organizations, as well as specialists who operate equipment aboard ships of the merchant marine and river fleets and the Ministry of the Fishing Industry.

This handbook can also be used by students of higher educational institutions and technical schools as well as students of marine training institutions in their course and graduation design.

#### Foreword

The requirements for navigational safety are increasing significantly in connection with the growth and tonnage and increases in the speed, size and inertia of modern vessels. The radio navigation instruments used aboard ships of the merchant marine and fishing industry fleets are extremely important. These

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instruments make it possible to reduce the number of accidents, which result in substantial material losses, and sometimes create a real threat to the environment.

Volume II of the Handbook on Shipboard Radio Communication and Radio Navigation Equipment is devoted to modern domestic radio navigation instruments used aboard ships of the merchant marine and fishing industry fleets. Examined here are functional and structural diagrams, operating and technical characteristics, and rules for technical operation of radio navigation instruments.

Chapter 1 examines shipboard radio direction finders with audible readout of the "Rybka" and "Barkas" type, and the type "Rumb" two-channel visual radio direction finder.

Chapter 2 contains a description of the types "Pirs-1D" and "Pirs-1M" receivers used to determine the location of the vessel by means of signals emitted by the shore stations of the "Dekka" phase radio navigation system.

Chapter 3 examines the type KPI-5F shipboard receiver, which is used to determine the location of the ship by means of signals from the shore stations of the pulse-phase "Loran-S" radio navigation system.

Chapters 4 and 5 provide descriptions of the type "Lotsiya", "Mius" and "Nayada" shipboard navigational RLS [radar stations] and type "Okean" and "Okean-M" automated navigational RLS.

Chapter 6 examines the operating features of the "Istra" navigational Doppler RLS for measuring docking speed.

Chapter 7 is devoted to the "Mgla" night-vision shipboard infrared equipment and the "Gorizont" television system.

Each chapter in the handbook is concluded with recommendations on the installation, mounting and operation of the radio navigation equipment in question aboard the ship.

The development of technical ship navigation equipment in the past ten years is characterized by the widespread introduction of digital computing technology and the use of new systems principles which increase the operational capabilities of instruments. The implementation of circuits for digital processing of radio navigation data makes it possible to use optimal processing methods and to represent navigational data in a form convenient for the navigator.

At the time work on the handbook was being finished, the development of the "Yenisey" radar set was completed and experimental operation had begun. The "Briz-Ye" and "Kron" computerized situation indicators are nearing completion, and production will begin in 1981. The more sophisticated "Biryusa" navigational set is now under development.

Unfortunately, none of these navigational sets are included in this issue of the Handbook. However, the authors have noted the trends in the development of

77

technical ship navigation equipment, and have tried to provide a more thorough explanation of those questions which will help the reader to overcome difficulties in studying the radar-computing units not covered in the present Handbook.

This work was done by a collective of authors. Chapter 1 was written by 0. V. Kononov, Chapter 2, by A. V. Zherlakov, Chapter 3 by Yu. Ye. Gornostayev, Chapters 4 and 6 by A. M. Bayrashevskiy, Chapters 5 and 7 by N. T. Nichiporenko, and Section 4.2 by A. A. Il'in.

Please direct any remarks to 191065, Leningrad, ul. Gogolya, 8, izdatel'stvo "Sudostroyeniye".

#### Table of Contents

Forewo	ord	3
Part :	I. Shipboard Radio Navigation Systems	5
Chapte	er 1. Shipboard Direction-Finding Radio Navigation System	8
1.1.	Shipboard radio direction finders and their classification	8
1.2.	"Rybka" shipboard radio direction finder with audible indication	9
1.3.	"Barkas" portable shipboard radio direction finder with audible indication	14
1.4.	"Rumb" two-channel shipboard visual radio direction finder	18
1.5.	Requirements for installing, adjusting and operating shipboard radio direction finders	28
Chapt	er 2. Phase-Type Radio Navigation Systems	30
2.1.	Classification and features of shipboard phase-type radio navigation systems	30
2.2.	"Pirs-1D" shipboard receiver display	36
2.3.	"Pirs-1M" shipboard receiver display	50
2.4.	Requirements for installation, adjustment and operation of receiver display under shipboard conditions	56
Chapt	er 3. Pulse and Pulse-Phase Radio Navigation Systems	59
3.1.	Classification and features of shipboard pulse and pulse-phase radio navigation systems	59
3.2.	KPI-5F receiver display	63

# FOR OFFICIAL USE ONLY

3.3. Recommendations for shipboard installation of receiver display	77
3.4. Basic operating rules for KPI-5F receiver display	77
Part II. Shipboard Navigational Radar Stations	85
Chapter 4. Pulsed Shipboard Navigational RLS	87
4.1. Features of pulsed RLS and their operational and technical characteristics	87
4.2. "Lotsiya" shipboard navigational RLS	95
4.3. "Mius" shipboard navigational RLS	L06
4.4. "Nayada"-series shipboard navigational RLS	L21
4.5. Requirements for installation, adjustment and operation of shipboard navigational RLS	L39
Chapter 5. Automated Shipboard Navigational RLS 1	L <b>52</b>
5.1. "Okean" shipboard navigational RLS	L52
5.2. "Okean-M" shipboard navigational RLS	176
5.3. Features of shipboard radio navigational and radar sets 2	210
Chapter 6. Doppler Navigational RLS 2	214
6.1. Operating features of Doppler navigational RLS 2	214
6.2. "Istra" RLS for measuring docking speed of vessels 2	218
Chapter 7. Shipboard Infrared and Television Equipment 2	221
7.1. "Mgla" infrared night-vision equipment 2	221
7.2. "Gorizont" shipboard television installation 2	25
Bibliography	29
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79

UDC 621.317.38

# MEASUREMENT OF ELECTRIC POWER IN THE AUDIO FREQUENCY RANGE

Leningrad IZMERENIYE ELEKTRICHESKOY MOSHCHNOSTI V ZVUKOVOM DIAPAZONE CHASTOT in Russian 1980 (signed to press 25 Oct 80) pp 2, 166-7

[Annotation and table of contents from book "Measurement of Electric Power in the Audio Frequency Range", by Aleksandra Yakovlevna Bezikovich and Yefim Zinov'yevich Shapiro, Izdatel'stvo "Energiya", 5,900 copies, 168 pages]

[Text] This book examines modern methods and means of measuring the true power in alternating current circuits in the power and audio frequency ranges. Block diagrams are presented for analog and digital wattmeters and devices for calibrating power-measuring equipment. Information on this subject from domestic and foreign literature is likewise catalogued.

This book is intended for power-measurement specialists engaged in the development, production and operation of measurement equipment. It will also be useful for metrologists involved in the calibration of various types of power-measurement equipment and for students at higher education establishments of the electrical engineering professions.

	Contents					Page
Foreword	i	•	•	•		3
Chapter	1. Aspects of the Theory of Electric Power Measurement					5
1.1.	Power in direct and alternating current circuits	•	•		•	5
1.2.	Direct and indirect measurement of power	•	•	•	•	. 7
1.3.	Method of direct estimation and the method of comparison v	whe	n			
	measuring power		•	•	•	12
1.4.	Measuring power by the simultaneous comparison method .	•	•		٠	19
1.5.	Methods of comparing equal and unequal values		•	•	•	21
1.6.	Methods of constructing measuring power transducers .	•	•	•	•	24
Chanter	2. Direct-Evaluation Instruments for Measurement of True	Po	wer	•		34
2.1.						34
	Electrostatic wattmeters					45
	Wattmeters with Hall-type transducers					49
	Wattmeters with modulation measuring power transducers	•	•	•	•	51

Chapter 3.1.	3. Comparison Instruments for Measuring Electric Power Types of measuring power transducers for realizing the comparison	62
3.1.	method	62
3.2.	Comparison instruments with electromechanical measuring power	
	transducers	72
3.3.	Electrothermal transducers for comparison instruments	77
3.4.	Thermoelectrical comparison instruments with passive input value	
	adders	81
3.5.	Thermoelectrical comparison instruments with active input value	
	adders	90
Chapter	4. Construction Features of Digital Wattmeters for Alternating	
-	Current	105
4.1.	General information	105
4.2.	Digital wattmeters with automatic error correction of measuring	
	power transducers	108
4.3.	Digital wattmeters with partial analog transduction of the input	
	values	113
4.4.	Digital microwattmeter with additive correction of methodical error	119
4.5.	Digital wattmeter based on the statistical testing method	122
4.6.	Digital wattmeters with transduction of power into an interval of	142
4.0.	time or into frequency	125
	time of into frequency	14.
Chapter	5. Primary Transducers of Voltage and Current	130
5.1.	Types of primary transducers and their basic parameters	130
5.2.	Resistive primary voltage transducers	133
5.3.	Resistive primary current transducers	137
5.4.	Resistors of primary voltage and current transducers	139
Chanter	6. Calibration of Power-Measurement Equipment	145
6.1.	Calibration features of power-measurement equipment	145
6.2.	Complete calibration of power-measurement equipment	146
6.3.	Experimentally calculated methods of calibrating power-measurement	140
0.5.	equipment	153
	equipment	133
Bibliog	caphy	162

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9512

CSO: 1860/242

UDC 621.396.6

#### PRECISION DESIGN IN MICROELECTRONICS

Leningrad TOCHNOSTNYYE RASCHETY V MIKROELEKTRONIKE in Russian 1980 (signed to press 5 Feb 80) pp 2, 138-39

[Annotation and table of contents from book "Precision Design in Miroelectronics" by Anatoliy Semenovich Shumilin, Izdatel'stvo Leningradskogo universiteta, 3099 copies, 140 pages]

[Text] This book analyzes the requirements for the accuracy of the output parameter of the microelectronic device as a complex system and presents detailed studies on the influence of the instabilities of technological influences on its accuracy. The author examines the most popular variants of technological processes: vacuum-thermal: ion-plasma, and semiconductor processes. As a result of mathematical studies, the author found calculated relations for the most characteristic points of accuracy losses.

The book is intended for a broad section of readers and primarily for researchers and engineers working in the area of microelectronics, as well as for senior students in these areas.

Contents	Page
Foreword	4
Chapter 1. Technological Accuracy of Microelectronic Systems	6
1.1. Equation of Technological Accuracy	6
1.2. Precision Design of Complex Systems	10
1.3. Synthesis of Optimal Tolerances for Individual Operations of the Technological Process	20
Chapter 2. Precision Design in Thin-Film Technology	35
2.1. Block Diagram of the Technological Process and the Composition of Calculations	35
2.2. Basic Elements of Thin-Film Microcircuits and Sources of Errors	27
in Their Manufacturing	37
2.3. Accuracy of Specific Volume Resistance of the Film	41
2.4. Accuracy of the Thickness and Specific Resistance of the Square	
of the Film	46
2.5. Accuracy of Achieving the Dielectric Constant	54

#### FOR OFFICIAL USE ONLY

2.6.	Accuracy of Achieving Geometrical Dimensions by Photolighographic Methods	55
2.7.	Influence of the Shading and Evaporation Effects and Errors in	٦.
2.0	the Basing of Masks in the Masking Method	70
2.8.	Special Characteristics of Precision Design in Using Ion Technology	77
	3. Precision Design in Semiconductor Technology	79
3.1.	Block Diagram of the Technological Process and the Composition	
3.2.	of Calculations Component Parts of Semiconductor Microcircuits and Their Basic	79
	Parameters	82
3.3.	Accuracy of Achieving the Parameters of the Epitaxial Layer	92
3.4.	Accuracy of Achieving the Parameters by the Technological Diffusion Process	100
3.5.	Accuracy in Manufacturing Oxide Masking Films on a Semiconductor	102 113
Chapter A	4. Selection of the Means of Technological Control	116
4.1.	Classification of the Means of Controlling Integrated	115
	Microcircuits	115
4.2.	Evaluation of Effectiveness and Selection of a Complex of	
	the Means of Control by Statistical Methods	121
Supplemen	nt. Computation of Optimal Tolerances on an Electronic Computer	
•	by the Method of Dynamic Programming	131
3ibliogra	aphy	134
J	• •	174

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10,233

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#### RELIABILITY OF MULTICHANNEL COMMUNICATIONS SYSTEMS

Moscow NADEZHNOST' SISTEM MNOGOKANAL'NOY SVYAZI in Russian 1980 (signed to press 20 Aug 80) pp 2, 94-95

[Annotation and table of contents from book "Reliability of Multichannel Communications Systems", by Ivan Ivanovich Gnidenko and Nikolay Parfent'yevich Truskalov, Izdatel srvo "Svyaz'", 6000 copies, 96 pages]

[Text] Questions connected with determining the basic indicators of the reliability of primary system components and factors affecting the equipment during its production and use are discussed. Methods for designing and insuring reliability of the primary system, its equipment and transmission systems are described. Questions concerning inspection, preventive maintenance, organization of repairs and providing spare parts for electrocommunications offices are examined. Principles for designing systems for assembling information about the reliability of the equipment and installations of the primary system and methods for processing statistical data on their failures are presented.

The book is for specialists studying questions of reliability during planning, production and operation of multichannel communications systems, as well as for students of electrical engineering communications institutes.

CONTENTS	Page
Foreword	3
Introduction	5
Chapter 1. Insuring reliability of the primary system 1.1 Indicators of the reliability of primary system components 1.2 Methods for insuring reliability of the primary system 1.3 External factors causing failures of primary system components	8 8 15 20
Chapter 2. Methods for calculating reliability during planning and operation of systems and transmission lines of the primary system  2.1 Selection and basis for reliability indicators during planning	29 <b>2</b> 9

84

in t	culation of reliability of transmission systems equipment the developmental stage	39
trai	ermination of certain reliability indicators of nsmission systems equipment based on performance data	41
	nomical effectiveness of improving transmission systems iability of the primary system	44
	ethods for providing reliability of transmission systems nd of the primary system while in operation	46
	gram to insure reliability of transmission systems equipment ucture of the primary system maintenance work system	46 49
4.1 Sup	upplying spare parts for the primary system plying electrocommunications offices with complete sets spare parts, tools and accessories for repair and with	68
com	plete sets of replaceable parts ucture for supplying spare units to a sector of the	68
prin	mary system imum distribution of the quantity of spare units in the	72
str	ucture of a sector of the primary system	76
	nformation about primary system equipment and	79
5.1 Sour	rces of information on the primary system equipment installation reliability	79
5.2 Sign	nificance of reliability information for operational erprises of the primary system	80
5.3 Asse	embly, distribution and realization of information about mary system equipment and installation reliability	81
	The parameter $\Delta$ of the Poisson distribution given various values of the index (1- $\beta$ , 1- $\epsilon_2$ , $\epsilon_1$ )	87
Appendix 2. V	Values of coefficient R <sub>1</sub>	88
Appendix 3. \	Values of coefficient R <sub>2</sub>	89
2	Initial data and formulae for calculating the probability of the lack of spare units at OUP [attended tandem station], ZRB [zone repair center] and TsRB [central repair center]	
·	warehouses	89
	Formulas for calculating coefficients $1^{\Delta A}$ , $2^{\Delta A}$ , $3^{\Delta A}$	91
	Form for recording equipment failures of transmission systems	91
Bibliography		92
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9194 CSO: 1860/212	· 2	

85

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SYNTHESIS OF REDUNDANT DISCRETE SYSTEMS WITH STRUCTURAL RECONFIGURATION

Kiev SINTEZ IZBYTOCHNYKH DISKRETNYKH USTROYSTV S REKONFIGURATSIYEY STRUKTURY in Russian 1979 (signed to press 3 May 79) pp 2, 155-6

[Annotation and table of contents from the book "The Synthesis of Redundant Discrete Systems with Structural Reconfiguration", by Vadim Yevgen'yevich Obukhov and Vadim Vladimirovich Pavlov, Izdatel'stvo "Naukova dumka", 1500 copies, 156 pages]

[Text] This monograph constitutes an examination of the method which incorporates structure-reliability synthesis of redundant discrete systems (DS) subject to internal disturbances(failures). The method has its basis in the development of the prescriptive theory of dynamic systems as it applies to DS as a class. The applied body of mathematics provides the means for synthesis of redundant structures of diagnosible DS with continuous self-recovery of predetermined functional characteristics. Boolean functional equations of absolute invariance constitute the bases of this body of mathematics. A description of the DS reconfiguration method is given. Also provided is an overview of the synthesis of redundant diagnosible and autodiagnosible DS with adaptive recovery.

This book is intended for the use of scientific research workers and postgraduate students specializing in the field of computer technology, as well as for engineers involved in reliability design of discrete systems.

#### Table of Contents

Foreword	
Chapter 1	
General Considerations of Reliability in Discrete Systems	5
1.1. Definitions of reliability theory	á
1.2. Methods for increasing reliability 1.3. Analysis of some reliability problems	8

86

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Chapt	er 2	
Struc	tural Redundancy in Discrete Systems	13
2.1. 2.2. 2.3.	Parallel reservation of noncontrollable DS Reservation of controllable DS Homogeneous DS with variable structure	13 20 22
Chapt	er 3	
Funct	ional Synthesis of Systems	31
3.1. 3.2. 3.3. 3.4. 3.5.	Dynamic systems and automatic devices Failures and disturbances Prescriptive theory of dynamic system synthesis Functional Boolean equations of absolute invariance Functional synthesis of structures in finite automatic devices	31 36 42 43 46
Chapt	er 4	
Recon	afiguration of Redundant Discrete Systems	48
4.1. 4.2. 4.3.	Principles of reconfiguration An example of redundant DS with structural reconfiguration Functional and structural features of the structural reconfiguration method Reconfiguration of redundant DS with autocontrol	48 51 55 61
Chapt	er 5	
Mathe	ematical Bases for the Synthesis of Redundant Discrete Systems Structural Reconfiguration	<b>6</b> 8
5.1. 5.2. 5.5.5.6. 7.8.	Boolean equations and analysis of methods for their solution Solution of Boolean equations with distributed terms Solution of Boolean equations using the Veitch diagram Special features of the preceding solutions General algorithm for the solution of extended Boolean equations Generalized algorithm for the solution of Boolean equations Matrix method for the solution of Boolean equations Solution of Boolean equations of arbitrary structure	68 79 87 89 93 102 106
Chapt	ter 6	
	ability Synthesis of Logic Systems	116
6.1. 6.2.	Synthesis of redundant structure in a reverse-code converter Synthesis of redundant structure in a converter of direct	116
6.3.	substitution code to 2421 code	120

87

Chapt	er 7						
Relia	ability Synthesis of Automatic Devices with Memory	137					
-	Characteristic features of the description of Boolean equations for automatic devices with memory	137					
7.2. General synthesis of redundant structure in automatic devices with memory							
7.3. Synthesis of redundant structure in counters and scaling circuits							
7.4.	.4. Simultaneous synthesis of redundant structure in logic devices and automatic devices with memory						
Bibli	ogr <b>a</b> phy	152					
COPYF	RIGHT: Izdatel'stvo "Naukova dumka", 1979						
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#### THEORY AND CALCULATION OF OPTOELECTRONIC INSTRUMENTS

Moscow TEORIYA I RASCHET OPTIKO-ELEKTRONNYKH PRIBOROV in Russian 1980 pp 2-6

[Annotation and table of contents from book "Theory and Calculation of Optoelectronic Instruments", by Yu. G. Yakushenkov, Second edition, revised and supplemented, Izdatel'stvo "Sovetskoye radio", 392 pages]

[Text] The physical basics of optoelectronic instruments are presented. Typical components of the primary information processing system of an optoelectronic instrument are described: optical radiation sources and receivers, optical system, image analyzers, scanning systems and modulators. Basic attention is devoted to the calculation and selection of the parameters and characteristics of these components from the standpoint of optimum mating with each other and signal filtration on a background of interference.

Procedures are presented for calculating and selecting the most important parameters of optoelectronic instruments: range, precision, signal to noise ratio, design parameters and characteristics of sources, optical systems, receivers and electronic circuitry.

The book is intended as a textbook for students enrolled in optical courses at colleges and universities. It may also be beneficial to a wide range of specialists, engaged in optical instrument design and familiar with physics and electronics within the scope of the programs of general engineering institutions.

The text is accompanied by 86 figures, 23 tables and 131 bibliographic references.

		Contents	Page
Pr	eface		7
1.	Intr	oduction	10
	1.1.	Optoelectronic Instruments: Definition, Generalized Diagram and	
		Operating Methods	10
	1.2.	Brief Classification of Optoelectronic Instruments	13
		Comparison of Optoelectronic Instruments with Visual Optical and	
		with Radio Electronic Instruments	14
	1.4.	Brief Historical Outline on Development of Optoelectronic Instrument	
		Design	17

89

Par	t I.	Physical Fundamentals and Elements of Optoelectronic Instruments	
2.	Onti	cal Radiation. Sources of Radiation	21
	2.1.	Optical Spectrum of Electromagnetic Waves	21
	2.2.	Basic Energy and Photometric Parameters and Relations Between Them	22
	2.3.	System of Astrophysical Celestial Values	31
	2.4.	Basic Parameters and Characteristics of Radiators. Black Body	33
	2.5.	Laws of Thermal Radiation	37
	2.6.	Incandescent Lamps and Gas Discharge Tubes	42
	2.7.	Light-Emitting Diodes and Electroluminescent Panels and Capacitors Application of Lasers as Radiation Sources in Optoelectronic	45
		Instruments	48
3.	Infl	uence of Propagation Medium of Optical Radiation on Performance of	
	Opto	electronic Instruments	58
	3.1.	General Aspects of Propagation of Radiation in Atmosphere	58
	3.2.	Absorption of Radiation in Earth's Atmosphere	62
	3.3.	Scattering of Radiation in Atmosphere	66
	3.4.	Fluctuations of Atmospheric Transparency and Their Effect on	
	•••	Performance of Optoelectronic Instrument	72
	3.5.	Refraction of Optical Rays	75
4.	Onti	cal System of Optoelectronic Instrument	77
•	4.1.	Purpose and Features of Optical System of Optical Calculations of Optical Systems of Optical Calculations of Optical Systems of Optical Systems of Optical Calculations	77
		tronic Instruments. Basic Energy Equation of Optoelectronic	
		Instrument	79
	4.3.	Energy Losses in Optical System of Optoelectronic Instrument	91
	4.4.	Image Quality as Function of Parameters of Optical System	92
	4.5.	Transmitting Optical Systems of Optoelectronic Instruments	96
	4.6.	Lenses of Optoelectronic Instruments	99
	4.7.	Capacitors of Receiving Optical Systems of Optoelectronic Instruments	104
	4.8.	Optical Analyzers	111
		Optical Correctors	119
		Blinds	125
	4.11.	Optical Filters	128
	4.12.	Materials of Optical Systems of Optoelectronic Instruments	134
5.	Radi	ation Receiver as Component of Optoelectronic Instrument	140
	5.1.	Classification of Radiation Receivers	140
	5.2.	Parameters of Radiation Receivers	142
	5.3	Characteristics of Radiation Receivers	150
	5.4	Certification of Receivers. Conversion of Their Parameters	153
	5.5	Cooling of Radiation Receivers	159
	5.6	Basic Kinds of Radiation Receivers and Their Application in Opto- electronic Instruments	161
	c 7	Application of Coordinate (Position-Sensitive) Radiation Receivers	
		in Optoelectronic Instruments	176
	5.8	Matching of Receiver with Radiation Source and Optical System of Optoelectronic Instrument	18
	5.9	Matching of Radiation Receiver with Electronic System of Optoelec-	190
		tronic Instrument	190

90

6.	6.1. 6.2.	ning in Optoelectronic Instruments Purpose and Role of Scanning Parameters and Characteristics of Scanning Systems Basic Types of Scanning Systems	195 195 197 201
Pa:	rt II	. Theoretical Fundamentals of Signal Reception in Application to ctronic Instruments	
۰p		COUNTY INSCIDENCES	
7.	Sign	als and Interference in Optoelectronic Instruments	204
	7.1.	Deterministic Signals and Methods of Describing Them	204
	7.2.	Random Signals and Methods of Describing Them	215
	7.3.	Quality Criteria of Optoelectronic Instruments	219
8.	Modu 8.1.	lation and Demodulation of Signal in Optoelectronic Instrument Purpose, Classification and Features of Modulation of Radiation	222
		Flux in Optoelectronic Instrument	222
		Signal Power Losses in Modulation	233
	8.3.	Optical Signal Modulation Using Rasters	235
	8.4.	Electrooptical and Certain Other Types of Modulators	241
	8.5.	Demodulation of Signal in Optoelectronic Instrument	244
9.	Opto	electronic Instrument as Linear Filter	248
	9.1.	Structural Diagram of Optoelectronic Tracking System	248
	9.2.	Optical System as Linear Filter (Fourier Transformation in Incoherent Optical System)	252
	9 3	Fourier Transformation in Coherent Optical System	252
	9 4	Space-Frequency Characteristic of Raster of Analyzer	259
		Spectrum of Deterministic Signal at Output of Primary Information	263
		Processing System	267
	9.6.	Passage of Random Interference Through Optoelectronic System	271
	9.7.	Transfer Function of Propagation Medium	273
10.	Opt	ical Signal Reception on Interference Background	277
]	10.1.	Methods of Receiving Optical Signals	277
:	10.2.	Optimal Filtration in Signal Detection on Interference Background	286
1	10.3.	Optimal Filtration in Signal Parameter Measurement	293
]	10.4.	Spectral Filtration	299
]	10.5.	Spatial Filtration in Incoherent Optical Systems	304
	10.6.	Spatial Filtration in Coherent Optical Systems	313
]	10.7.	Optical Correlation in Optoelectronic Instruments	317
Paı Opt	rt III	I. Methods of Calculating Basic Parameters and Characteristics of ctronic Instruments	
11	Dwo	and the Calculation and Danis Tourist Co. 1	
H.		codure for Calculating and Designing Typical Optoelectronic	700
1		Features of Design and Development of Optoelectronic Instruments	322
1	1.2.	Basic Calculations Done in Design of Optoelectronic Instruments	322 325
		. <del>-</del>	
12.	Ener	rgy (Optical Engineering) Calculations of Optoelectronic Instruments	326
1	2.1.	Generalized Energy Calculation Procedure	326
1	2.2.	Calculation of Signal Detection Probability on Interference and Noise Background	333

91

## FOR OFFICIAL USE ONLY

12.3. Efficiency Calculation of Optoelectronic Instrument	335
12.4. Calculation of Signal-to-Noise Ratio at Radiation Receiver Output	338
12.5. Calculation of Effective Range of Optoelectronic Instrument	340
13. Precision Calculations of Optoelectronic Instruments	344
13.1. Basic Stages of Precision Calculation of Optoelectronic Instrument	344
13.2. Potential Precision Calculation of Optoelectronic Instrument	346
13.3. Calculation of Dynamic Errors for Deterministic Input Signals	348
13.4. Calculation of Fluctuation Errors for Stationary Random Signals	349
13.5. Calculation of Instrumental Errors	351
14. Calculation of Certain Design Parameters of Optoelectronic Instruments	352
·14.1. Calculation of Diameter of Input Lens of Optoelectronic Instrument	352
14.2. Calculation of Field of View of Optical System	354
14.3. Calculation of Basic Parameters of Radiation Sources	357
14.4. Selection and Calculation of Basic Parameters of Radiation Receiver	360
14.5. Selection and Calculation of Parameters of Scanning System	364
15. Calculation and Selection of Certain Dynamic Parameters of Optoelectronic	:
Instruments	368
15.1. Comparative Evaluation and Selection of Kind of Optical Signal	
Modulation	368
15.2. Selection of Working Modulation Frequencies	374
15.3. Selection of Passband of Electronic Circuit	377
Bibliography	381
Subject Index	388

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#### THEORY OF CODE DIVISION OF SIGNALS

Moscow TEORIYA KODOVOGO RAZDELENIYA SIGNALOV in Russian 1980 (signed to press 3 Dec 80) pp 2-4, 207-8

[Annotation, excerpts from foreword and table of contents from book "Theory of Code Division of Signals", by Igor' Mikhaylovich Pyshkin, Izdatel'stvo "Svyaz'", 4,000 copies, 208 pages]

[Excerpts] This book presents the basis of the theory of code multiplexing and division of signals as applied to asynchronous address systems for the transmission of information. An analysis of the performance in terms of error probability for the most widely used types of asynchronous address systems is carried out for given models of interference. The synthesis and analysis of the best possible structures for processing signals in asynchronous address systems are also carried out. This book investigates performance in terms of error probability for nonparametric detectors and examines the feasibility of transmitting continuous traffic in asynchronous address systems using nonparametric modulation methods. This book likewise studies the use of code multiplexing and division of signals in practicable systems for the transmission of information.

This book is intended for engineering and technical workers in the field of data transmission.

#### Foreword

The networks and systems of data transmission that now exist or are in the design stage must insure simultaneous communications for a great many fixed and mobile installations randomly disposed over a given territory. As a result of this, multistation and multichannel data-transmission systems, the design of which requires a comprehensive approach, form the basis of future communications networks. Along with the multistation and multichannel capabilities of modern data-transmission systems, their basic features are:

- the transmission of both continuous and digital information on the same communication channels;
- the introduction of pulse modulation;
- the application of digital methods of shaping and processing signals;

93

- the application of equipment based on digital integrated circuits.

Considering the overloading of the frequency range and the great cost of cable lines for the construction of multichannel and multistation systems, it is necessary to develop and introduce methods for multiplexing and dividing signals which make it possible to improve performance in terms of error probability and the traffichandling capability of communication networks.

The development of new operating principles for multichannel and multistation systems should improve their traffic-handling capacity, increase their immunity to industrial interference and jamming and insure reliable operation under conditions of multipath radio-signal propagation and electromagnetic compatibility (EMC) of various radiotechnical systems.

All the positive characteristics of code multiplexing and division, based on the utilization of the redundancy of complex signals, manifest themselves most fully in asynchronous multistation data transmission systems or asynchronous address systems (AAS's). At the same time, the problems solved during the construction of AAS's extend beyond the scope of those problems relating to the analysis and synthesis of a specific system. The basic feature of AAS operation which distinguishes these systems from multistation systems with frequency and time-division of signals is the presence of structured interference, that is, interference whose wave form is coincident with the signal and which operates in the same band as the useful signal and which overlaps the signal with respect to time. The detection of a useful (working) signal of a given AAS subscriber is carried out against the background of signals (interfering signals) from the system's other subscribers. These signals overlap with respect to time and frequency. The influence of the interfering signals at the detector input of any of the system subscribers creates interference which can be approximated by a random process whose distribution density is essentially different from Gaussian. As a result of this, the optimum linear circuits for processing signals against a background of white Gaussian noise as well as the methods for the optimum processing of signals in the presence of correlated Gaussian noise gives unsatisfactory results in this case.

The main task for the theory of code multiplexing and division of signals is the task of "separating" the signals, that is, the selection of criteria and the discovery of algorithms that make it possible under the influence of non-Gaussian noise to detect and separate the useful signals and measure their parameters.

										Con	ten	ts											Page
Foreword	i.													•					•		•		3
Introduc	tion	•				•	•	•	•	•	•	.•	•	•	•	•	•	•	•	•	•	•	6
Chapter	1 /	70 <b>70</b>	<b>-</b> 1	Сħ	ara	cte	rie	tic	9 (	of A	svn	chi	onc	านธ	Ado	ires	s S	yst	ems	fo	r		
-		Dа	ta	Tra	n sn	iss	ion					•	•	•	•	•	•	•	•	•	•		12
1.1.	Class	aifi	cat	ion	of	mu	ılti	acc	ess	S	rste	ems			•	•	•	•	•	•	•	•	12
1.2	Dioi	ra1	asv	nch	ron	ous	ad	dre	82	sy 8	sten	ns f	or	dat	ta 1	tran	smi	ssi	on	•	•	•	22
1.3.	Repr	esen	tat	ion	ar	id s	ele	cti	.on	of	sig	gna]	.s i	or	as	yncr	iron	ous	ao	are	:88		24
	SV	sten	ıs					•	•	•	•	•	٠	•	•	•	•	٠	•	•	•	•	24

Chapter 2. Performance in Terms of Error Probability and the Efficiency	
of Asynchronous Address Systems When Normalizing Mutual	
Interference	. 32
2.1. Energy and correlation methods of describing mutual interference .	
2.2. Correlation method of determining mutual interference	. 35
2.3. Efficiency of asynchronous address systems when transmitting	
digital information	. 40
· ·	
Chapter 3. Performance in Terms of Error Probability When Using	
Time-and-Frequency Signals in a Channel with Constant	
Parameters	. 42
3.1. General characteristics of asynchronous address systems with	
time-and-frequency signals	. 42
3.2. Performance in terms of error probability for asynchronous address	
systems with time-and-frequency signals	. 48
systems with time-and-frequency signals	• 40
3.3. Performance in terms of error probability for coherent detection	
of digital information exposed to structured interference	. 50
3.4. Performance in terms of error probability for asynchronous address	
systems during coherent detection and individual processing of	•
the elements	. 59
3.5. Performance in terms of error probability for asynchronous address	*
systems during coherent detection by sequence estimation	. 63
Chapter 4. Performance in Terms of Error Probability for Asynchronous	
Address Systems with Time-and-Frequency Signals and Ran-	
domly Varying Phase	. 74
4.1. Introductory remarks	74
4.2. Performance in terms of error probability for noncoherent detec-	
tion of digital information exposed to structured interference .	. 76
4.3. Performance in terms of error probability for asynchronous address	
systems with time-and-frequency signals during noncoherent	
symbol-by-symbol processing	. 87
	. 0,
4.4. Performance in terms of error probability for asynchronous address	0.0
systems with quadratic summation of the signal elements	. 92
4.5. Performance in terms of error probability for the noncoherent	
detection of time-and-frequency signals in asynchronous address	
systems with coherent summation of the elements	. 101
ol	
Chapter 5. Questions of the Design and Optimal Structures for	
Asynchronous Address Systems	. 110
5.1. Stating the problem of design	. 110
5.2. Design for the optimal detector of address signal elements in	
non-Gaussian interference	. 115
5.3. Realizing a quasi-optimal detector of time-and-frequency signal	
elements for asynchronous address systems	. 129
5.4. Investigating performance in terms of error probability for	
optimal detectors	. 133
5.5. Asymptotically optimal method of weight accumulation	. 149
5.5. Asymptotically optimal method of weight decimalation	15/

# . 95

e - M		ress sy																	•	•
6.2.		ving th																		
6.3.		sis and																		
		digit																		
		modula																		
Bibli	ograph	у	• .	•	•	•	•	•	•	٠	•	•	•	•	•	•	•	•	•	. •
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TRANSMISSION OF IMAGES IN DIGITAL FORM

Moscow PEREDACHA IZOBRAZHENIY V TSIFROVOY FORME in Russian 1980 (signed to press 23 Jul 80) pp 3-4, 120

[Foreword and table of contents from book "Transmission of Images in Digital Form", by Boris Yevseyevich Trofimov and Oleg Valentinovich Kulikovskiy, Izdatel'stvo "Svyaz'", 5900 copies, 120 pages]

[Text]

FOREWORD

The modern stage in the development of communications technology is characterized by the widespread introduction of digital image transmission systems. Such systems are more efficient than analog systems. The high efficiency of digital systems is explained by three factors: the possibility of regenerating the digital signals in intermediate and terminal equipment, the high technical and economic indicators of equipment for digital signal processing and correction and the possibility for using existing digital communications networks for transmission of television, video telephone and facsimilie images in digital form.

An almost total liberation of a digital signal from linear distortions and interference occurs in the regeneration process. This permits images to be transmitted over communications lines with a large number of regenerators over a distance which is practically unlimited. During transmission of television programs via Earth satellites and for digital image transmission from space it is possible to reduce the power of on-board transmitters as the result of digital encoding. Regeneration and encoding permit multiple re-recording of images and noticeably to improve the equipment indices which are used for correction of aperture and geometrical distortions, color control, combined image synthesis, analysis of initial images, conversion of signals from one standard to another and for other ends.

There are about 100 methods presently known for converting the analog signal of an image into a digital one. The extent of their suitability is determined by the concrete operating conditions of the system. All of the methods are based on one of the following types of modulation: pulse code (IKM), differential pulse code (DIKM) and delta modulation (DM). Moreover, the methods differ in the type of processing of the analog signal before analog-to-digital modulation and the type of subsequent processing of the initial digital signal. The properties of the image and of the visual analyzer are taken into consideration, which permits the maximal efficiency of the signal conversion process to be attained overall.

97

Recently several books and topical collections devoted to digital image transmission methods have been published in our country and abroad. Theoretical questions of conversion of images into digital form based primarily on IKM are examined in [1]. Certain theoretical aspects of orthogonal image conversion are examined in [2]. A number of articles of a review nature are published in [3-5]. A number of questions associated with conversion of television images into digital form have found reflection in [6].

In this book a description and analysis of the methods for analog-to-digital conversion of semitone images are given in systematized form, taking into consideration the peculiarities of their transmission via communication channels.

The authors acknowledge advice and critical remarks expressed by Prof. and Doctor of Technical Sciences N. B. Zeliger, honored scientist and technician of the RSFSR, Doctor of Technical Sciences Ye. L. Orlovskiy and Candidate of Technical Sciences A. G. Likiardopulo, and also the great task of reviewing and editing the book performed by V. S. Ignatkin and Candidate of Technical Sciences Yu. M. Braude-Zolotarev.

We ask that wishes and remarks concerning the book be directed to the "Svyaz'" publishing house at the address 101000, Moscow Chistoprudnyy Blvd., 2.

The Authors

	CONTENTS	Page
Foreword		3
Chapter 1.	Image transmission and reproduction	5
	Electric signals for image transmission	5 9
	The image transmission channel	
1.3	Peculiarities of sensing and image quality evaluation criteria	11
Chapter 2	. Digitization of signals	21
2.1	Principles of digitization	21
2.2	Distortion during digitization	25
	Distortion during restoration	28
Chanter 3	. Analog-to-digital conversion based on IKM (pulse code	
Glapter 5	modulation)	33
3 1	Pulse code modulation	33
• • • •	Precode signal processing	37
	Postcode unit-by-unit processing of digital signals	47
3.4		63
Chanter 4	. Analog-to-digital conversion based on delta modulation	71
4.1	Delta modulation with singular integrator	71
4.2	Delta modulation with double integrator	78
	Adaptive delta modulation	82
4.7	Markette serve measures.	

Chapter 5	Analog-to-digital conversion	on based on differential	
	pulse code modulation (DIK	and the second s	88
5.1 Differential pulse code modulation with single integrator			88
5.2 Refined prediction with DIKM			95
	Improved DIKM algorithms		101
5.4	Differential pulse code modul	ation with irregular metering	
	currents	•	107
Conclusion		111	
			115
Bibliography		113	
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9194			
	)/218	END	